- High-Performance Fixed-Point Digital
 Signal Processors (DSPs) TMS320C62x™
 - 5-, 4-, 3.33-ns Instruction Cycle Time
 - 200-, 250-, 300-MHz Clock Rate
 - Eight 32-Bit Instructions/Cycle
 - 1600, 2000, 2400 MIPS
- C6202 and C6203B GLS Ball Grid Array (BGA) Packages are Pin-Compatible With the C6204 GLW BGA Package[†]
- C6202B and C6203B GNZ and GNY Packages are Pin-Compatible
- VelociTI[™] Advanced Very-Long-Instruction-Word (VLIW) C62x[™] DSP Core
 - Eight Highly Independent Functional Units:
 - Six ALUs (32-/40-Bit)
 - Two 16-Bit Multipliers (32-Bit Result)
 - Load-Store Architecture With 32 32-Bit General-Purpose Registers
 - Instruction Packing Reduces Code Size
 - All Instructions Conditional
- Instruction Set Features
 - Byte-Addressable (8-, 16-, 32-Bit Data)
 - 8-Bit Overflow Protection
 - Saturation
 - Bit-Field Extract, Set, Clear
 - Bit-Counting
 - Normalization
- 3M-Bit On-Chip SRAM
 - 2M-Bit Internal Program/Cache (64K 32-Bit Instructions)
 - 1M-Bit Dual-Access Internal Data (128K Bytes)
 - Organized as Two 64K-Byte Blocks for Improved Concurrency
- 32-Bit External Memory Interface (EMIF)
 - Glueless Interface to Synchronous Memories: SDRAM or SBSRAM
 - Glueless Interface to Asynchronous Memories: SRAM and EPROM
 - 52M-Byte Addressable External Memory Space

- Four-Channel Bootloading Direct-Memory-Access (DMA) Controller With an Auxiliary Channel
- Flexible Phase-Locked-Loop (PLL) Clock Generator
- 32-Bit Expansion Bus (XBus)
 - Glueless/Low-Glue Interface to Popular PCI Bridge Chips
 - Glueless/Low-Glue Interface to Popular Synchronous or Asynchronous Microprocessor Buses
 - Master/Slave Functionality
 - Glueless Interface to Synchronous FIFOs and Asynchronous Peripherals
- Three Multichannel Buffered Serial Ports (McBSPs)
 - Direct Interface to T1/E1, MVIP, SCSA Framers
 - ST-Bus-Switching Compatible
 - Up to 256 Channels Each
 - AC97-Compatible
 - Serial-Peripheral Interface (SPI)
 Compatible (Motorola™)
- Two 32-Bit General-Purpose Timers
- IEEE-1149.1 (JTAG[‡]) Boundary-Scan-Compatible
- 352-Pin BGA Package (GJL) (C6202)
- 352-Pin BGA Package (GNZ) (C6202B)
- 384-Pin BGA Package (GLS) (C6202)
- 384-Pin BGA Package (GNY) (C6202B)
- 0.18-μm/5-Level Metal Process (C6202)
 0.15-μm/5-Level Metal Process (C6202B)
 - CMOS Technology
- 3.3-V I/Os, 1.8-V Internal (C6202)
 3.3-V I/Os, 1.5-V Internal (C6202B)



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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Motorola is a trademark of Motorola, Inc.

Other trademarks are the property of their respective owners.

† For more details, see the GLS BGA package bottom view.

[‡] IEEE Standard 1149.1-1990 Standard-Test-Access Port and Boundary Scan Architecture.

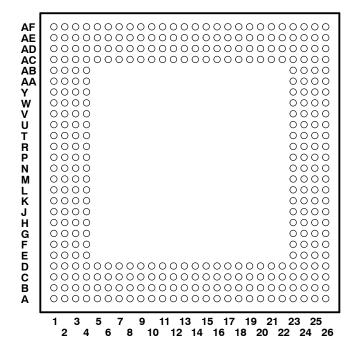


SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

GJL, GNZ, GLS, and GNY BGA packages 3 description 5 device characteristics 6 C62x device compatibility 8 functional and CPU (DSP core) block diagram 9 CPU (DSP core) description 10 memory map summary 12 peripheral register descriptions 13 DMA synchronization events 18 interrupt sources and interrupt selector 19 signal groups description 20 signal descriptions 23 development support 34 documentation support 37 clock PLL 37 power-down mode logic 40 power-supply sequencing 43 IEEE 1149.1 JTAG compatibility statement 45 absolute maximum ratings over operating case temperature ranges 46	parameter measurement information
recommended operating conditions	JTAG test-port timing 98 mechanical data 99 revision history 103

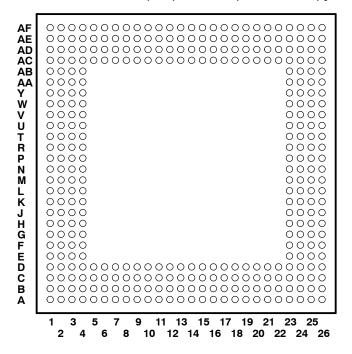
GJL, GNZ, GLS, and GNY BGA packages

GJL 352-PIN BALL GRID ARRAY (BGA) PACKAGE (BOTTOM VIEW) [C6202 only]

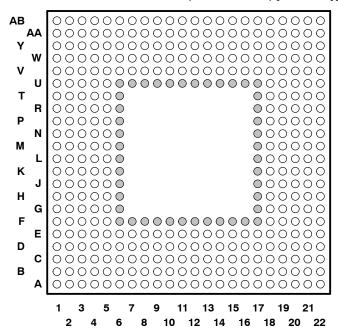


GJL, GNZ, GLS, and GNY BGA packages (continued)

GNZ 352-PIN BALL GRID ARRAY (BGA) PACKAGE (BOTTOM VIEW) [C6202B only]



GLS 384-PIN BGA PACKAGE (BOTTOM VIEW) [C6202 only]



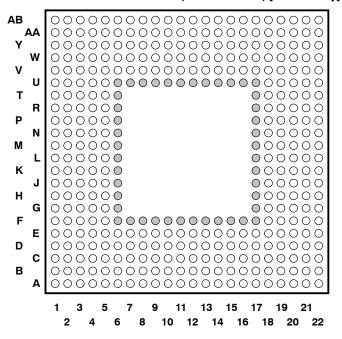
The C6202 and C6203B GLS BGA packages are pin-compatible with the C6204 GLW package except that the inner row of balls (which are additional power and ground pins) are removed for the C6204 GLW package.

These balls are NOT applicable for the C6204 devices 340-pin GLW BGA package.



GJL, GNZ, GLS, and GNY BGA packages (continued)

GNY 384-PIN BGA PACKAGE (BOTTOM VIEW) [C6202B only]



description

The TMS320C6202 and TMS320C6202B devices are part of the TMS320C62x[™] fixed-point DSP generation in the TMS320C6000[™] DSP platform. The C62x[™] DSP devices are based on the high-performance, advanced VelociTI[™] very-long-instruction-word (VLIW) architecture developed by Texas Instruments (TI), making these DSPs an excellent choice for multichannel and multifunction applications.

The TMS320C62x™ DSP offers cost-effective solutions to high-performance DSP-programming challenges. The TMS320C6202/02B has a performance capability of up to 2400 million instructions per second (MIPS) at 300 MHz. The C6202/02B DSP possesses the operational flexibility of high-speed controllers and the numerical capability of array processors. These processors have 32 general-purpose registers of 32-bit word length and eight highly independent functional units. The eight functional units provide six arithmetic logic units (ALUs) for a high degree of parallelism and two 16-bit multipliers for a 32-bit result. The C6202/02B can produce two multiply-accumulates (MACs) per cycle. This gives a total of 600 million MACs per second (MMACS) for the C6202/02B device. The C6202/02B DSP also has application-specific hardware logic, on-chip memory, and additional on-chip peripherals.

The C6202/02B devices program memory consists of two blocks, with a 128K-byte block configured as memory-mapped program space, and the other 128K-byte block user-configurable as cache or memory-mapped program space. Data memory for the C6202/02B consists of two 64K-byte blocks of RAM.

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SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

description (continued)

The C6202/02B device has a powerful and diverse set of peripherals. The peripheral set includes three multichannel buffered serial ports (McBSPs), two general-purpose timers, a 32-bit expansion bus (XBus) that offers ease of interface to synchronous or asynchronous industry-standard host bus protocols, and a glueless 32-bit external memory interface (EMIF) capable of interfacing to SDRAM or SBSRAM and asynchronous peripherals.

The C62x[™] devices have a complete set of development tools which includes: a new C compiler, an assembly optimizer to simplify programming and scheduling, and a Windows[™] debugger interface for visibility into source code execution.

device characteristics

Table 1 provides an overview of the TMS320C6202, TMS320C6202B, TMS320C6203B, and the TMS320C6204 pin-compatible DSPs. The table shows significant features of each device, including the capacity of on-chip RAM, the peripherals, the execution time, and the package type with pin count, etc. This data sheet primarily focuses on the functionality of the TMS320C6202/02B devices although it also identifies to the user the pin-compatibility of the C6202 and C6203B GLS, and the C6204 GLW BGA packages. This data sheet also identifies the pin-compatibility of the C6202B and the C6203B GNZ and GNY packages. For the functionality information on the TMS320C6203B device, see the *TMS320C6203B Fixed-Point Digital Signal Processor* data sheet (literature number SPRS086). For the functionality information on the TMS320C6204 device, see the *TMS320C6204 Fixed-Point Digital Signal Processor* data sheet (literature number SPRS152). And for more details on the C6000™ DSP device part numbers and part numbering, see Table 16 and Figure 4.

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device characteristics (continued)

Table 1. Characteristics of the Pin-Compatible DSPs

HARDV	VARE FEATURES	C6202	C6202B	C6203B	C6204
	EMIF	√	V	V	√
Peripherals	DMA	4-Channel	4-Channel With Throughput Enhancements	4-Channel With Throughput Enhancements	4-Channel With Throughput Enhancements
	Expansion Bus	√	$\sqrt{}$	$\sqrt{}$	√
	McBSPs	3	3	3	2
	32-Bit Timers	2	2	2	2
	Size (Bytes)	256K	256K	384K	64K
Internal Program Memory	Organization	Block 0: 128K-Byte Mapped Program Block 1: 128K-Byte Cache/Mapped Program	Block 0: 128K-Byte Mapped Program Block 1: 128K-Byte Cache/Mapped Program	Block 0: 256K-Byte Mapped Program Block 1: 128K-Byte Cache/Mapped Program	1 Block: 64K-Byte Cache/Mapped Program
	Size (Bytes)	128K	128K	512K	64K
Internal Data Memory	Organization	2 Blocks: Four 16-Bit Banks per Block 50/50 Split	2 Blocks: Four 16-Bit Banks per Block 50/50 Split	2 Blocks: Four 16-Bit Banks per Block 50/50 Split	2 Blocks: Four 16-Bit Banks per Block 50/50 Split
CPU ID + CPU Rev ID	Control Status Register (CSR.[31:16])	0x0002	0x0003	0x0003	0x0003
Frequency	MHz	200, 250	250, 300	250, 300	200
Cycle Time	ns	4 ns (6202-250) 5 ns (6202-200)	3.33 ns (6202B-300) 4 ns (6202B-250) 4 ns (02BGNZA-250)	3.33 ns (6203B-300) 4 ns (6203B-250) 4 ns (03BGNZA-250)	5 ns (6204-200)
Voltage	Core (V)	1.8	1.5	1.5 1.7	1.5
_	I/O (V)	3.3	3.3	3.3	3.3
PLL Options	CLKIN frequency multiplier [Bypass (x1), x4, x6, x7, x8, x9, x10, and x11]	x1, x4 (Both Pkgs)	All PLL Options (GNY Pkg) x1, x4, x8, x10 (GNZ Pkg)	All PLL Options (GLS/GNY Pkgs) x1, x4, x8, x10 (GNZ Pkg)	x1, x4 (Both Pkgs)
	27 x 27 mm	352-pin GJL	352-pin GNZ	352-pin GNZ	-
DO 4	18 x 18 mm	384-pin GLS	-	384-pin GLS	340-pin GLW
BGA Packages	18 x 18 mm	-	384-pin GNY	384-pin GNY (2.x, 3.x only)	-
	16 x 16 mm	-	-	-	288-pin GHK
Process Technology	μт	0.18 μm	0.15 μm	0.15 μm	0.15 μm
Product Status [†]	Product Preview (PP) Advance Information (AI) Production Data (PD)	PD	PD	PD	PD

[†] PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.



SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

C62x[™] device compatibility

The TMS320C6202, C6202B, C6203B, and C6204 devices are pin-compatible; thus, making new system designs easier and providing faster time to market. The following list summarizes the C62x DSP device characteristic differences:

Core Supply Voltage (1.8 V versus 1.7 V versus 1.5 V)

The C6202 device core supply voltage is 1.8 V while the C6202B, C6203B, C6204 devices have core supply voltages of 1.5 V. Furthermore, the C6203B-300 speed devices (GNY and GNZ packages) also have a 1.7-V core supply voltage.

Device Clock Speeds

The C6202B and C6203B devices run at -250 and -300 MHz clock speeds (with a C620xBGNZA extended temperature device that also runs at -250 MHz), while the C6202 device runs at -200 and -250 MHz, and the C6204 device runs at -200 MHz clock speed.

PLL Options Availability

Table 1 identifies the available PLL multiply factors [e.g., CLKIN x1 (PLL bypassed), x4, etc.] for each of the C62x DSP devices. For additional details on the PLL clock module and specific options for the C6202/02B devices, see the Clock PLL section of this data sheet.

For additional details on the PLL clock module and specific options for the C6203B device, see the Clock PLL section of the *TMS320C6203B Fixed-Point Digital Signal Processor Data Sheet* (literature number SPRS086).

And for additional details on the PLL clock module and specific options for the C6204 device, see the *Clock PLL* section of the *TMS320C6204 Fixed-Point Digital Signal Processor Data Sheet* (literature number SPRS152).

On-Chip Memory Size

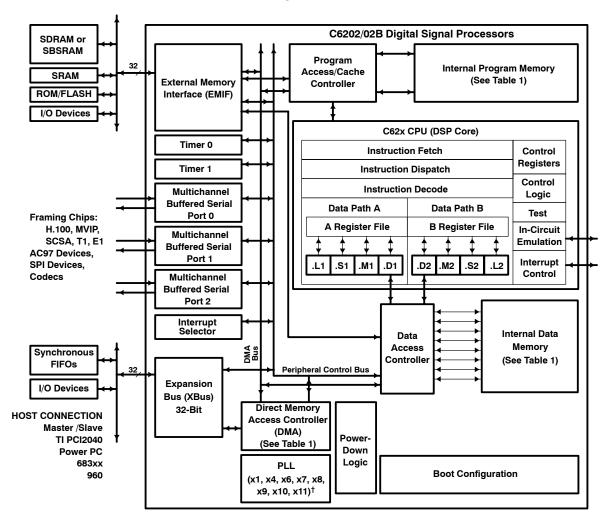
The C6202/02B, C6203B, and C6204 devices have different on-chip program memory and data memory sizes (see Table 1).

McBSPs

The C6202, C6202B, and C6203B devices have three McBSPs while the C6204 device has two McBSPs on-chip.

For a more detailed discussion on migration concerns, and similarities/differences between the C6202, C6202B, C6203B, and C6204 devices, see the *How to Begin Development Today and Migrate Across the TMS320C6202/02B/03B/04 DSPs Application Report* (literature number SPRA603).

functional and CPU (DSP core) block diagram



[†] For additional details on the PLL clock module and specific options for the C6202/02B devices, see Table 1 and the Clock PLL section of this data sheet.

SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

CPU (DSP core) description

The CPU fetches VelociTI advanced very-long instruction words (VLIW) (256 bits wide) to supply up to eight 32-bit instructions to the eight functional units during every clock cycle. The VelociTI VLIW architecture features controls by which all eight units do not have to be supplied with instructions if they are not ready to execute. The first bit of every 32-bit instruction determines if the next instruction belongs to the same execute packet as the previous instruction, or whether it should be executed in the following clock as a part of the next execute packet. Fetch packets are always 256 bits wide; however, the execute packets can vary in size. The variable-length execute packets are a key memory-saving feature, distinguishing the C62x CPU from other VLIW architectures.

The CPU features two sets of functional units. Each set contains four units and a register file. One set contains functional units .L1, .S1, .M1, and .D1; the other set contains units .D2, .M2, .S2, and .L2. The two register files each contain 16 32-bit registers for a total of 32 general-purpose registers. The two sets of functional units, along with two register files, compose sides A and B of the CPU [see the functional and CPU (DSP core) block diagram and Figure 1]. The four functional units on each side of the CPU can freely share the 16 registers belonging to that side. Additionally, each side features a single data bus connected to all the registers on the other side, by which the two sets of functional units can access data from the register files on the opposite side. While register access by functional units on the same side of the CPU as the register file can service all the units in a single clock cycle, register access using the register file across the CPU supports one read and one write per cycle.

Another key feature of the C62x CPU is the load/store architecture, where all instructions operate on registers (as opposed to data in memory). Two sets of data-addressing units (.D1 and .D2) are responsible for all data transfers between the register files and the memory. The data address driven by the .D units allows data addresses generated from one register file to be used to load or store data to or from the other register file. The C62x CPU supports a variety of indirect addressing modes using either linear- or circular-addressing modes with 5- or 15-bit offsets. All instructions are conditional, and most can access any one of the 32 registers. Some registers, however, are singled out to support specific addressing or to hold the condition for conditional instructions (if the condition is not automatically "true"). The two .M functional units are dedicated for multiplies. The two .S and .L functional units perform a general set of arithmetic, logical, and branch functions with results available every clock cycle.

The processing flow begins when a 256-bit-wide instruction fetch packet is fetched from a program memory. The 32-bit instructions destined for the individual functional units are "linked" together by "1" bits in the least significant bit (LSB) position of the instructions. The instructions that are "chained" together for simultaneous execution (up to eight in total) compose an execute packet. A "0" in the LSB of an instruction breaks the chain, effectively placing the instructions that follow it in the next execute packet. If an execute packet crosses the 256-bit-wide fetch-packet boundary, the assembler places it in the next fetch packet, while the remainder of the current fetch packet is padded with NOP instructions. The number of execute packets within a fetch packet can vary from one to eight. Execute packets are dispatched to their respective functional units at the rate of one per clock cycle and the next 256-bit fetch packet is not fetched until all the execute packets from the current fetch packet have been dispatched. After decoding, the instructions simultaneously drive all active functional units for a maximum execution rate of eight instructions every clock cycle. While most results are stored in 32-bit registers, they can be subsequently moved to memory as bytes or half-words as well. All load and store instructions are byte-, half-word, or word-addressable.

CPU (DSP core) description (continued)

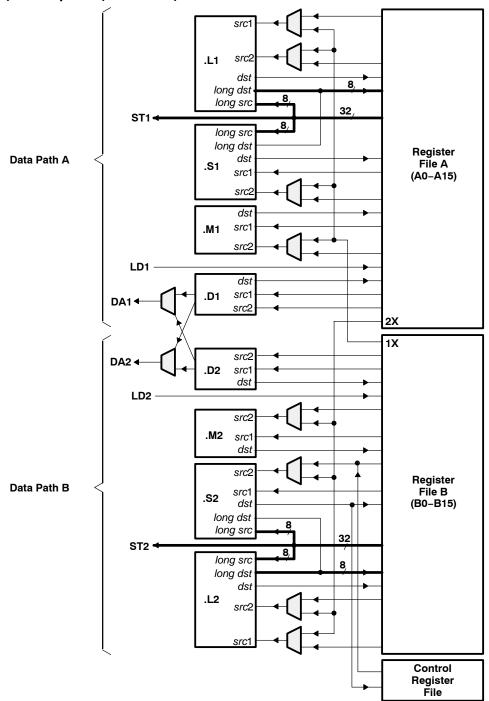


Figure 1. TMS320C62x CPU (DSP Core) Data Paths

memory map summary

Table 2 shows the memory map address ranges of the C6202/02B device. The C6202/02B device has the capability of a MAP 0 or MAP 1 memory block configuration. These memory block configurations are set up at reset by the boot configuration pins (generically called BOOTMODE[4:0]). For the C6202/02B device, the BOOTMODE configuration is handled, at reset, by the expansion bus module (specifically XD[4:0] pins). For more detailed information on the C6202/02B device settings, which include the device boot mode configuration at reset and other device-specific configurations, see TMS320C620x/C670x DSP Boot Modes and Configuration (literature number SPRU642).

Table 2. TMS320C6202/02B Memory Map Summary

MEMORY BLOCK DESCRIPTION		BLOCK SIZE	HEV ADDRESS DANSE
MAP 0	MAP 1	(BYTES)	HEX ADDRESS RANGE
External Memory Interface (EMIF) CE0	Internal Program RAM	256K	0000_0000-0003_FFFF
EMIF CE0	Reserved	4M-256K	0004_0000-003F_FFFF
EMIF CE0	EMIF CE0	12M	0040_0000-00FF_FFFF
EMIF CE1	EMIF CE0	4M	0100_0000-013F_FFFF
Internal Program RAM	EMIF CE1	256K	0140_0000-0143_FFFF
Reserved	EMIF CE1	4M-256K	0144_0000-017F_FFFF
EMIF Regis	sters	256K	0180_0000-0183_FFFF
DMA Controller F	Registers	256K	0184_0000-0187_FFFF
Expansion Bus (XBu	us) Registers	256K	0188_0000-018B_FFFF
McBSP 0 Reg	gisters	256K	018C_0000-018F_FFFF
McBSP 1 Reg	gisters	256K	0190_0000-0193_FFFF
Timer 0 Regi	isters	256K	0194_0000-0197_FFFF
Timer 1 Regi	isters	256K	0198_0000-019B_FFFF
Interrupt Selector Registers		512	019C_0000-019C_01FF
Power-Down Registers		256K-512	019C_0200-019F_FFFF
Reserved		256K	01A0_0000-01A3_FFFF
McBSP 2 Registers		256K	01A4_0000-01A7_FFFF
Reserve	d	5.5M	01A8_0000-01FF_FFFF
EMIF CE	2	16M	0200_0000-02FF_FFFF
EMIF CE	3	16M	0300_0000-03FF_FFFF
Reserve	d	1G-64M	0400_0000-3FFF_FFF
XBus XCE0			4000_0000-4FFF_FFF
XBus XCE1		256M	5000_0000-5FFF_FFF
XBus XCE2		256M	6000_0000-6FFF_FFF
XBus XCE3		256M	7000_0000-7FFF_FFF
Internal Data	RAM	128K	8000_0000-8001_FFFF
Reserve	d	2G–128K	8002_0000-FFFF_FFF

peripheral register descriptions

Table 3 through Table 13 identify the peripheral registers for the C6202/02B device by their register names, acronyms, and hex address or hex address range. For more detailed information on the register contents, bit names, and their descriptions, see the peripheral reference guide referenced in *TMS320C6000 DSP Peripherals Overview Reference Guide* (literature number SPRU190).

Table 3. EMIF Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0180 0000	GBLCTL	EMIF global control	
0180 0004	CECTL1	EMIF CE1 space control	External or internal; dependent on MAP0 or MAP1 configuration (selected by the MAP bit in the EMIF GBLCTL register)
0180 0008	CECTL0	EMIF CE0 space control	External or internal; dependent on MAP0 or MAP1 configuration (selected by the MAP bit in the EMIF GBLCTL register)
0180 000C	-	Reserved	
0180 0010	CECTL2	EMIF CE2 space control	Corresponds to EMIF CE2 memory space: [0200 0000-02FF FFFF]
0180 0014	CECTL3	EMIF CE3 space control	Correspond to EMIF CE3 memory space: [0300 0000–03FF FFFF]
0180 0018	SDCTL	EMIF SDRAM control	
0180 001C	SDTIM	EMIF SDRAM refresh control	
0180 0020 – 0180 0054	_	Reserved	
0180 0058 – 0183 FFFF	_	Reserved	

Table 4. DMA Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
0184 0000	PRICTL0	DMA channel 0 primary control
0184 0004	PRICTL2	DMA channel 2 primary control
0184 0008	SECCTL0	DMA channel 0 secondary control
0184 000C	SECCTL2	DMA channel 2 secondary control
0184 0010	SRC0	DMA channel 0 source address
0184 0014	SRC2	DMA channel 2 source address
0184 0018	DST0	DMA channel 0 destination address
0184 001C	DST2	DMA channel 2 destination address
0184 0020	XFRCNT0	DMA channel 0 transfer counter
0184 0024	XFRCNT2	DMA channel 2 transfer counter
0184 0028	GBLCNTA	DMA global count reload register A
0184 002C	GBLCNTB	DMA global count reload register B
0184 0030	GBLIDXA	DMA global index register A
0184 0034	GBLIDXB	DMA global index register B
0184 0038	GBLADDRA	DMA global address register A
0184 003C	GBLADDRB	DMA global address register B
0184 0040	PRICTL1	DMA channel 1 primary control
0184 0044	PRICTL3	DMA channel 3 primary control



SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

peripheral register descriptions (continued)

Table 4. DMA Registers (Continued)

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME
0184 0048	SECCTL1	DMA channel 1 secondary control
0184 004C	SECCTL3	DMA channel 3 secondary control
0184 0050	SRC1	DMA channel 1 source address
0184 0054	SRC3	DMA channel 3 source address
0184 0058	DST1	DMA channel 1 destination address
0184 005C	DST3	DMA channel 3 destination address
0184 0060	XFRCNT1	DMA channel 1 transfer counter
0184 0064	XFRCNT3	DMA channel 3 transfer counter
0184 0068	GBLADDRC	DMA global address register C
0184 006C	GBLADDRD	DMA global address register D
0184 0070	AUXCTL	DMA auxiliary control register
0184 0074-0187 FFFF	_	Reserved

Table 5. Expansion Bus (XBUS) Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0188 0000	XBGC	Expansion bus global control register	
0188 0004	XCECTL1	XCE1 space control register	Corresponds to XBus XCE0 memory space: [4000 0000–4FFF FFFF]
0188 0008	XCECTL0	XCE0 space control register	Corresponds to XBus XCE1 memory space: [5000 0000–5FFF FFFF]
0188 000C	XBHC	Expansion bus host port interface control register	DSP read/write access only
0188 0010	XCECTL2	XCE2 space control register	Corresponds to XBus XCE2 memory space: [6000 0000–6FFF FFFF]
0188 0014	XCECTL3	XCE3 space control register	Corresponds to XBus XCE3 memory space: [7000 0000–7FFF FFFF]
0188 0018	_	Reserved	
0188 001C		Reserved	
0188 0020	XBIMA	Expansion bus internal master address register	DSP read/write access only
0188 0024	XBEA	Expansion bus external address register	DSP read/write access only
0188 0028 - 018B FFFF		Reserved	
_	XBISA	Expansion bus internal slave address	
_	XBD	Expansion bus data	1

peripheral register descriptions (continued)

Table 6. Interrupt Selector Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
019C 0000	MUXH	Interrupt multiplexer high	Selects which interrupts drive CPU interrupts 10–15 (INT10–INT15)
019C 0004	MUXL	Interrupt multiplexer low	Selects which interrupts drive CPU interrupts 4–9 (INT04–INT09)
019C 0008	EXTPOL	External interrupt polarity	Sets the polarity of the external interrupts (EXT_INT4-EXT_INT7)
019C 000C-019C 01FF	_	Reserved	
019C 0200	PDCTL	Peripheral power-down control register	
019C 0204-019F FFFF	_	Reserved	

Table 7. Peripheral Power-Down Control Register

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	
019C 0200	PDCTL	Peripheral power-down control register	

Table 8. McBSP 0 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
018C 0000	DRR0	McBSP0 data receive register	The CPU and DMA controller can only read this register; they cannot write to it.
018C 0004	DXR0	McBSP0 data transmit register	
018C 0008	SPCR0	McBSP0 serial port control register	
018C 000C	RCR0	McBSP0 receive control register	
018C 0010	XCR0	McBSP0 transmit control register	
018C 0014	SRGR0	McBSP0 sample rate generator register	
018C 0018	MCR0	McBSP0 multichannel control register	
018C 001C	RCER0	McBSP0 receive channel enable register	
018C 0020	XCER0	McBSP0 transmit channel enable register	
018C 0024	PCR0	McBSP0 pin control register	
018C 0028-018F FFFF	_	Reserved	

SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

peripheral register descriptions (continued)

Table 9. McBSP 1 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0190 0000	DRR1	Data receive register	The CPU and DMA controller can only read this register; they cannot write to it.
0190 0004	DXR1	McBSP1 data transmit register	
0190 0008	SPCR1	McBSP1 serial port control register	
0190 000C	RCR1	McBSP1 receive control register	
0190 0010	XCR1	McBSP1 transmit control register	
0190 0014	SRGR1	McBSP1 sample rate generator register	
0190 0018	MCR1	McBSP1 multichannel control register	
0190 001C	RCER1	McBSP1 receive channel enable register	
0190 0020	XCER1	McBSP1 transmit channel enable register	
0190 0024	PCR1	McBSP1 pin control register	
0190 0028-0193 FFFF	_	Reserved	

Table 10. McBSP 1 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0190 0000	DRR1	Data receive register	The CPU and DMA controller can only read this register; they cannot write to it.
0190 0004	DXR1	McBSP1 data transmit register	
0190 0008	SPCR1	McBSP1 serial port control register	
0190 000C	RCR1	McBSP1 receive control register	
0190 0010	XCR1	McBSP1 transmit control register	
0190 0014	SRGR1	McBSP1 sample rate generator register	
0190 0018	MCR1	McBSP1 multichannel control register	
0190 001C	RCER1	McBSP1 receive channel enable register	
0190 0020	XCER1	McBSP1 transmit channel enable register	
0190 0024	PCR1	McBSP1 pin control register	
0190 0028-0193 FFFF	_	Reserved	

peripheral register descriptions (continued)

Table 11. McBSP 2 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
01A4 0000	DRR2	McBSP2 data receive register	The CPU and DMA controller can only read this register; they cannot write to it.
01A4 0004	DXR2	McBSP2 data transmit register	
01A4 0008	SPCR2	McBSP2 serial port control register	
01A4 000C	RCR2	McBSP2 receive control register	
01A4 0010	XCR2	McBSP2 transmit control register	
01A4 0014	SRGR2	McBSP2 sample rate generator register	
01A4 0018	MCR2	McBSP2 multichannel control register	
01A4 001C	RCER2	McBSP2 receive channel enable register	
01A4 0020	XCER2	McBSP2 transmit channel enable register	
01A4 0024	PCR2	McBSP2 pin control register	
01A4 0028-01A7 FFFF	_	Reserved	

Table 12. Timer 0 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0194 0000	CTL0	Timer 0 control register	Determines the operating mode of the timer, monitors the timer status, and controls the function of the TOUT pin.
0194 0004	PRD0	Timer 0 period register	Contains the number of timer input clock cycles to count. This number controls the TSTAT signal frequency.
0194 0008	CNT0	Timer 0 counter register	Contains the current value of the incrementing counter.
0194 000C-0197 FFFF	_	Reserved	

Table 13. Timer 1 Registers

HEX ADDRESS RANGE	ACRONYM	REGISTER NAME	COMMENTS
0198 0000	CTL1	Timer 1 control register	Determines the operating mode of the timer, monitors the timer status, and controls the function of the TOUT pin.
0198 0004	PRD1	Timer 1 period register	Contains the number of timer input clock cycles to count. This number controls the TSTAT signal frequency.
0198 0008	CNT1	Timer 1 counter register	Contains the current value of the incrementing counter.
0198 000C-019B FFFF	_	Reserved	

SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

DMA synchronization events

The C6202/C6202B DMA supports up to four independent programmable DMA channels. The four main DMA channels can be read/write synchronized based on the events shown in Table 14. Selection of these events is done via the RSYNC and WSYNC fields in the Primary Control registers of the specific DMA channel. For more detailed information on the DMA module, associated channels, and event-synchronization, see the TMS320C620x/C670x DSP Program and Data Memory Controller / Direct Memory Access (DMA) Comtroller Reference Guide (literature number SPRU577).

Table 14. TMS320C6202/02B DMA Synchronization Events

DMA EVENT NUMBER (BINARY)	EVENT NAME	EVENT DESCRIPTION	
00000	Reserved	Reserved	
00001	TINT0	Timer 0 interrupt	
00010	TINT1	Timer 1 interrupt	
00011	SD_INT	EMIF SDRAM timer interrupt	
00100	EXT_INT4	External interrupt pin 4	
00101	EXT_INT5	External interrupt pin 5	
00110	EXT_INT6	External interrupt pin 6	
00111	EXT_INT7	External interrupt pin 7	
01000	DMA_INT0	DMA channel 0 interrupt	
01001	DMA_INT1	DMA channel 1 interrupt	
01010	DMA_INT2	DMA channel 2 interrupt	
01011	DMA_INT3	DMA channel 3 interrupt	
01100	XEVT0	McBSP0 transmit event	
01101	REVT0	McBSP0 receive event	
01110	XEVT1	McBSP1 transmit event	
01111	REVT1	McBSP1 receive event	
10000	DSP_INT	Host processor-to-DSP interrupt	
10001	XEVT2	McBSP2 transmit event	
10010	REVT2	McBSP2 receive event	
10011 –11111	Reserved	Reserved. Not used.	

interrupt sources and interrupt selector

The C62x DSP core supports 16 prioritized interrupts, which are listed in Table 15. The highest-priority interrupt is INT_00 (dedicated to RESET) while the lowest-priority interrupt is INT_15. The first four interrupts (INT_00-INT_03) are non-maskable and fixed. The remaining interrupts (INT_04-INT_15) are maskable and default to the interrupt source specified in Table 15. The interrupt source for interrupts 4–15 can be programmed by modifying the selector value (binary value) in the corresponding fields of the Interrupt Selector Control registers: MUXH (address 0x019C0000) and MUXL (address 0x019C0004).

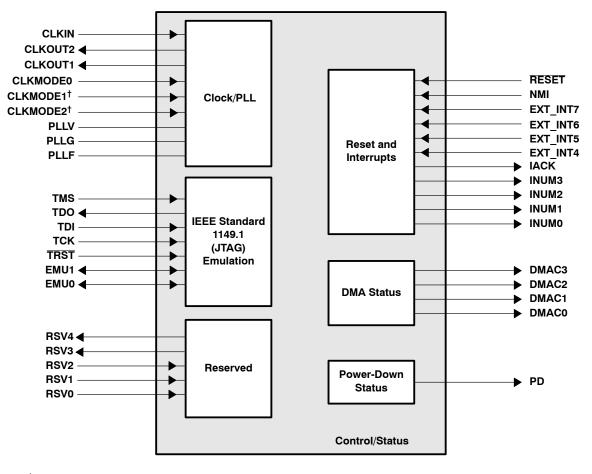
Table 15. C6202/02B DSP Interrupts

CPU INTERRUPT NUMBER	INTERRUPT SELECTOR CONTROL REGISTER	SELECTOR VALUE (BINARY)	INTERRUPT EVENT	INTERRUPT SOURCE
INT_00 [†]	_	_	RESET	
INT_01 [†]	_	_	NMI	
INT_02 [†]	_	_	Reserved	Reserved. Do not use.
INT_03 [†]	_	_	Reserved	Reserved. Do not use.
INT_04 [‡]	MUXL[4:0]	00100	EXT_INT4	External interrupt pin 4
INT_05 [‡]	MUXL[9:5]	00101	EXT_INT5	External interrupt pin 5
INT_06 [‡]	MUXL[14:10]	00110	EXT_INT6	External interrupt pin 6
INT_07 [‡]	MUXL[20:16]	00111	EXT_INT7	External interrupt pin 7
INT_08‡	MUXL[25:21]	01000	DMA_INT0	DMA channel 0 interrupt
INT_09 [‡]	MUXL[30:26]	01001	DMA_INT1	DMA channel 1 interrupt
INT_10 [‡]	MUXH[4:0]	00011	SD_INT	EMIF SDRAM timer interrupt
INT_11 [‡]	MUXH[9:5]	01010	DMA_INT2	DMA channel 2 interrupt
INT_12 [‡]	MUXH[14:10]	01011	DMA_INT3	DMA channel 3 interrupt
INT_13 [‡]	MUXH[20:16]	00000	DSP_INT	Host-processor-to-DSP interrupt
INT_14 [‡]	MUXH[25:21]	00001	TINT0	Timer 0 interrupt
INT_15 [‡]	MUXH[30:26]	00010	TINT1	Timer 1 interrupt
_	_	01100	XINT0	McBSP0 transmit interrupt
_	_	01101	RINT0	McBSP0 receive interrupt
		01110	XINT1	McBSP1 transmit interrupt
_		01111	RINT1	McBSP1 receive interrupt
		10000	Reserved	Reserved. Not used.
_		10001	XINT2	McBSP2 transmit interrupt
_		10010	RINT2	McBSP2 receive interrupt
_	_	10011 – 11111	Reserved	Reserved. Do not use.

 $^{^\}dagger$ Interrupts INT_00 through INT_03 are non-maskable and fixed.

[‡] Interrupts INT_04 through INT_15 are programmable by modifying the binary selector values in the Interrupt Selector Control registers fields. Table 15 shows the default interrupt sources for Interrupts INT_04 through INT_15. For more detailed information on interrupt sources and selection, see *TMS320C6000 DSP Interrupt Selector Reference Guide* (literature number SPRU646).

signal groups description



[†] CLKMODE1 and CLKMODE2 are NOT available on the C6202 device GJL package. CLKMODE2 is also NOT available on the GNZ package for the C6202B device.

Figure 2. CPU (DSP Core) Signals

signal groups description (continued)

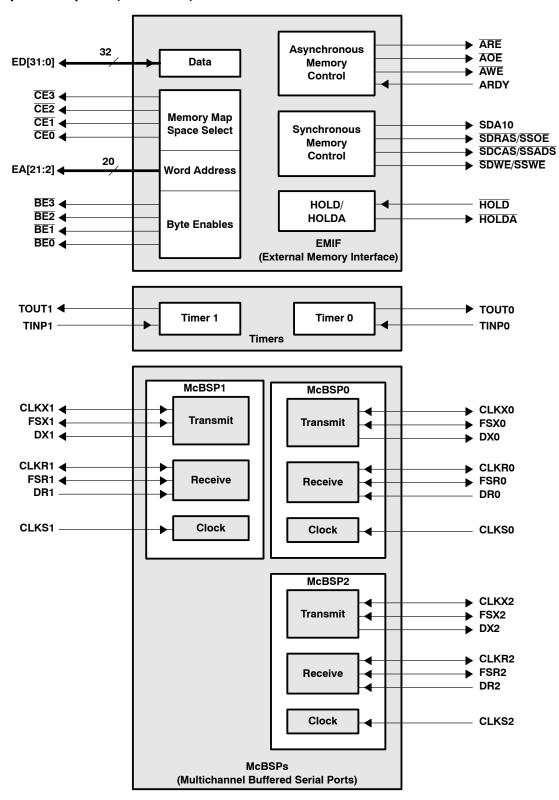


Figure 3. Peripheral Signals



signal groups description (continued)

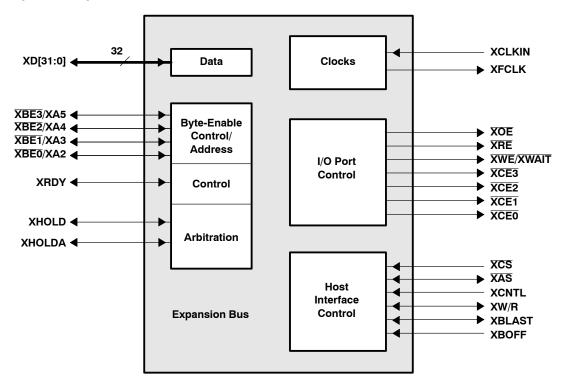


Figure 3. Peripheral Signals (Continued)

Signal Descriptions

	PIN	NO.						
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION				
	CLOCK/PLL							
CLKIN	C12	B10	I	Clock Input				
CLKOUT1	AD20	Y18	0	Clock output at full device speed				
CLKOUT2	AC10	AB19	0	Clock output at half (1/2) of device speed				
CLROU12	AC19	ADIS	0	☐ Used for synchronous memory interface				
CLKMODE0	B15	B12	I	Clock mode selects				
CLKMODE1	C11 [‡]	A9§	I	Selects what multiply factors of the input clock frequency the CPU frequency equals.				
CLKMODE2	-	A14 [§]	I	For more details on the GJL, GNZ, GLS, and GNY CLKMODE pins and the PLL multiply factors for the C6202 and C6202B devices, see the Clock PLL section of this data sheet.				
PLLV [¶]	D13	C11	A [#]	PLL analog V _{CC} connection for the low-pass filter				
PLLG [¶]	D14	C12	A [#]	PLL analog GND connection for the low-pass filter				
PLLF [¶]	C13	A11	A [#]	PLL low-pass filter connection to external components and a bypass capacitor				
				JTAG EMULATION				
TMS	AD7	Y5	I	JTAG test-port mode select (features an internal pullup)				
TDO	AE6	AA4	O/Z	JTAG test-port data out				
TDI	AF5	Y4	I	JTAG test-port data in (features an internal pullup)				
TCK	AE5	AB2	I	JTAG test-port clock				
TRST	AC7	AA3	I	JTAG test-port reset (features an internal pulldown)				
EMU1	AF6	AA5	I/O/Z	Emulation pin 1, pullup with a dedicated 20-k Ω resistor \parallel				
EMU0	AC8	AB4	I/O/Z	Emulation pin 0, pullup with a dedicated 20-k Ω resistor \parallel				
	_		-	RESET AND INTERRUPTS				
RESET	K2	J3	I	Device reset				
NMI	1.0	K2	I	Nonmaskable interrupt				
INIVII	L2	NΖ	Į į	☐ Edge-driven (rising edge)				
EXT_INT7	V4	U2		External interrupts				
EXT_INT6	Y2	U3] .	☐ Edge-driven				
EXT_INT5	AA1	W1	1	Polarity independently selected via the external interrupt polarity register bits				
EXT_INT4	W4	V2		(EXTPOL.[3:0])				
IACK	Y1	V1	0	Interrupt acknowledge for all active interrupts serviced by the CPU				
INUM3	V2	R3		Active interrupt identification number				
INUM2	U4	T1	0					
INUM1	V3	T2						
INUM0	W2	T3		☐ Encoding order follows the interrupt-service fetch-packet ordering				

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



[‡] For the C6202 GJL package only, the C11 pin is ground (V_{SS}). For all C62x™ GNZ packages, the C11 pin is CLKMODE1.

[§] For the C6202 GLS package, the CLKMODE2 (A14) and CLKMODE1 (A9) pins are internally unconnected.

PLLV, PLLG, and PLLF are not part of external voltage supply or ground. See the Clock PLL section for information on how to connect these pins.

[#] A = Analog Signal (PLL Filter)

 $[\]parallel$ For emulation and normal operation, pull up EMU1 and EMU0 with a dedicated 20-k Ω resistor. For boundary scan, pull down EMU1 and EMU0 with a dedicated 20-k Ω resistor.

	DIN	NO.		Signal Descriptions (Continued)
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE [†]	DESCRIPTION
				POWER-DOWN STATUS
PD	AB2	Y2	0	Power-down modes 2 or 3 (active if high)
				EXPANSION BUS
XCLKIN	A9	C8	1	Expansion bus synchronous host interface clock input
XFCLK	B9	A8	0	Expansion bus FIFO interface clock output
XD31	D15	C13		
XD30	B16	A13		
XD29	A17	C14		
XD28	B17	B14		
XD27	D16	B15		
XD26	A18	C15		
XD25	B18	A15		
XD24	D17	B16		
XD23	C18	C16		
XD22	A20	A17		Expansion bus data
XD21	D18	B17		Used for transfer of data, address, and control
XD20	C19	C17		Also controls initialization of DSP modes and expansion bus at reset [Note: For more information on pin control and boot configuration fields, see the Boot
XD19	A21	B18		
XD18	D19	A19		Modes and Configuration chapter of the TMS320C6000 DSP Peripherals Overview
XD17	C20	C18		Reference Guide (literature number SPRU190).]
XD16	B21	B19	1/0/7	XD[30:16] – XCE[3:0] memory type
XD15	A22	C19	I/O/Z	XD13 – XBLAST polarity
XD14	D20	B20		XD12 – XW/R polarity XD11 – Asynchronous or synchronous host operation
XD13	B22	A21		XD10 – Arbitration mode (internal or external)
XD12	E25	C21		XD9 – FIFO mode
XD11	F24	D20		XD8 – Little endian/big endian XD[4:0] – Boot mode
XD10	E26	B22		Noting Bootmode
XD9	F25	D21		All other expansion bus data pins not listed should be pulled down.
XD8	G24	E20		
XD7	H23	E21		
XD6	F26	D22		
XD5	G25	F20		
XD4	J23	F21		
XD3	G26	E22		
XD2	H25	G20		
XD1	J24	G21		
XD0	K23	G22		

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



OLONIAL	PIN NO.							
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION				
	EXPANSION BUS (CONTINUED)							
XCE3	F2	D2		Expansion bus I/O port memory space enables				
XCE2	E1	B1	0/7					
XCE1	F3	D3	O/Z					
XCE0	E2	C2		Only one asserted during any I/O port data access				
XBE3/XA5	C7	C5		Expansion bus multiplexed byte-enable control/address signals				
XBE2/XA4	D8	A4	1/0/7					
XBE1/XA3	A6	B5	I/O/Z	Act as byte-enable for host-port operation				
XBE0/XA2	C8	C6		Act as address for I/O port operation				
XOE	A7	A6	O/Z	Expansion bus I/O port output-enable				
XRE	C9	C7	O/Z	Expansion bus I/O port read-enable				
XWE/XWAIT	D10	B7	O/Z	Expansion bus I/O port write-enable and host-port wait signals				
XCS	A10	C9	I	Expansion bus host-port chip-select input				
XAS	D9	B6	I/O/Z	Expansion bus host-port address strobe				
XCNTL	B10	B9	I	Expansion bus host control. XCNTL selects between expansion bus address or data register.				
XW/R	D11	B8	I/O/Z	Expansion bus host-port write/read-enable. XW/R polarity is selected at reset.				
XRDY	A5	C4	I/O/Z	Expansion bus host-port ready (active low) and I/O port ready (active high)				
XBLAST	B6	B4	I/O/Z	Expansion bus host-port burst last-polarity selected at reset				
XBOFF	B11	A10	I	Expansion bus back off				
XHOLD	B5	A2	I/O/Z	Expansion bus hold request				
XHOLDA	D7	B3	I/O/Z	Expansion bus hold acknowledge				
		E	MIF - CON	ITROL SIGNALS COMMON TO ALL TYPES OF MEMORY				
CE3	AB25	Y21		Memory space enables				
CE2	AA24	W20	0/7					
CE1	AB26	AA22	O/Z	I — · ·				
CE0	AA25	W21		Only one asserted during any external data access				
BE3	Y24	V20		Byte-enable control				
BE2	W23	V21	O/Z	☐ Decoded from the two lowest bits of the internal address				
BE1	AA26	W22	0/2	Byte-write enables for most types of memory				
BE0	Y25	U20		Can be directly connected to SDRAM read and write mask signal (SDQM)				

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground

	PIN	NO		Signal Descriptions (Continued)
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION
	ı			EMIF - ADDRESS
EA21	J25	H20		
EA20	J26	H21		
EA19	L23	H22		
EA18	K25	J20	1	
EA17	L24	J21	1	
EA16	L25	K21	1	
EA15	M23	K20	1	
EA14	M24	K22	1	
EA13	M25	L21	1	
EA12	N23	L20]	Edward address (ward address)
EA11	P24	L22	O/Z	External address (word address)
EA10	P23	M20		
EA9	R25	M21		
EA8	R24	N22		
EA7	R23	N20		
EA6	T25	N21		
EA5	T24	P21		
EA4	U25	P20		
EA3	T23	R22		
EA2	V26	R21		
				EMIF - DATA
ED31	AD8	Y6		
ED30	AC9	AA6		
ED29	AF7	AB6		
ED28	AD9	Y 7		
ED27	AC10	AA7		
ED26	AE9	AB8		
ED25	AF9	Y8		
ED24	AC11	AA8]	
ED23	AE10	AA9	1/0/7	External data
ED22	AD11	Y9	I/O/Z	LAIGITIAI VAIA
ED21	AE11	AB10]	
ED20	AC12	Y10]	
ED19	AD12	AA10]	
ED18	AE12	AA11]	
ED17	AC13	Y11		
ED16	AD14	AB12		
ED15	AC14	Y12]	
ED14	AE15	AA12		

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



	PIN NO.			. g (
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION
				EMIF - DATA (CONTINUED)
ED13	AD15	AA13		
ED12	AC15	Y13		
ED11	AE16	AB13		
ED10	AD16	Y14		
ED9	AE17	AA14		
ED8	AC16	AA15		
ED7	AF18	Y15		
ED6	AE18	AB15	I/O/Z	External data
ED5	AC17	AA16		
ED4	AD18	Y16		
ED3	AF20	AB17		
ED2	AC18	AA17		
ED1	AD19	Y17		
ED0	AF21	AA18		
	1		EMIF	- ASYNCHRONOUS MEMORY CONTROL
ARE	V24	T21	O/Z	Asynchronous memory read-enable
AOE	V25	R20	O/Z	Asynchronous memory output-enable
AWE	U23	T22	O/Z	Asynchronous memory write-enable
ARDY	W25	T20	I	Asynchronous memory ready input
	EMIF - S'	YNCHRON	OUS DRA	AM (SDRAM)/SYNCHRONOUS BURST SRAM (SBSRAM) CONTROL
SDA10	AE21	AA19	O/Z	SDRAM address 10 (separate for deactivate command)
SDCAS/SSADS	AE22	AB21	O/Z	SDRAM column-address strobe/SBSRAM address strobe
SDRAS/SSOE	AF22	Y19	O/Z	SDRAM row-address strobe/SBSRAM output-enable
SDWE/SSWE	AC20	AA20	O/Z	SDRAM write-enable/SBSRAM write-enable
				EMIF – BUS ARBITRATION
HOLD	Y26	V22	I	Hold request from the host
HOLDA	V23	U21	0	Hold-request-acknowledge to the host
				TIMER 0
TOUT0	F1	D1	0	Timer 0 or general-purpose output
TINP0	H4	E2	I	Timer 0 or general-purpose input
				TIMER 1
TOUT1	J4	F2	0	Timer 1 or general-purpose output
TINP1	G2	F3	I	Timer 1 or general-purpose input
				DMA ACTION COMPLETE STATUS
DMAC3	Y3	V3		
DMAC2	AA2	W2		DMA called a constant
DMAC1	AB1	AA1	0	DMA action complete
DMAC0	AA3	W3		

DMAC0 A3 W3 L L

† I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

	PIN	NO.							
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION					
MULTICHANNEL BUFFERED SERIAL PORT 0 (McBSP0)									
CLKS0	M4	K3	I	External clock source (as opposed to internal)					
CLKR0	M2	L2	I/O/Z	Receive clock					
CLKX0	МЗ	K1	I/O/Z	Transmit clock					
DR0	R2	M2	I	Receive data					
DX0	P4	МЗ	O/Z	Transmit data					
FSR0	N3	M1	I/O/Z	Receive frame sync					
FSX0	N4	L3	I/O/Z	Transmit frame sync					
			MULT	ICHANNEL BUFFERED SERIAL PORT 1 (McBSP1)					
CLKS1	G1	E1	I	External clock source (as opposed to internal)					
CLKR1	J3	G2	I/O/Z	Receive clock					
CLKX1	H2	G3	I/O/Z	Transmit clock					
DR1	L4	H1	I	Receive data					
DX1	J1	H2	O/Z	Transmit data					
FSR1	J2	Н3	I/O/Z	Receive frame sync					
FSX1	K4	G1	I/O/Z	Transmit frame sync					
			MULT	ICHANNEL BUFFERED SERIAL PORT 2 (McBSP2)					
CLKS2	R3	N1	- 1	External clock source (as opposed to internal)					
CLKR2	T2	N2	I/O/Z	Receive clock					
CLKX2	R4	N3	I/O/Z	Transmit clock					
DR2	V1	R2	I	Receive data					
DX2	T4	R1	O/Z	Transmit data					
FSR2	U2	P3	I/O/Z	Receive frame sync					
FSX2	Т3	P2	I/O/Z	Transmit frame sync					
				RESERVED FOR TEST					
RSV0	L3	J2	- 1	Reserved for testing, pullup with a dedicated 20-k Ω resistor					
RSV1	G3	E3	I	Reserved for testing, pullup with a dedicated 20-k Ω resistor					
RSV2	A12	B11	I	Reserved for testing, pullup with a dedicated 20-k Ω resistor					
RSV3	C15	B13	0	Reserved (leave unconnected, <i>do not</i> connect to power or ground)					
RSV4	D12	C10	0	Reserved (leave unconnected, <i>do not</i> connect to power or ground)					

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground

CICNAL	PIN	NO.		eight Decemplians (Community)
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION
				SUPPLY VOLTAGE PINS
	A11	A3		
	A16	A7		
	B7	A16]	
	B8	A20]	
	B19	D4		
	B20	D6		
	C6	D7		
	C10	D9]	
	C14	D10]	
	C17	D13		
	C21	D14		
	G4	D16		
	G23	D17		
	Н3	D19		
	H24	F1]	
	K3	F4]	
	K24	F19		
	L1	F22		
	L26	G4		
DV_DD	N24	G19	S	3.3-V supply voltage (I/O)
	P3	J4]	
	T1	J19]	
	T26	K4]	
	U3	K19		
	U24	L1		
	W3	M22		
	W24	N4		
	Y4	N19		
	Y23	P4]	
	AD6	P19]	
	AD10	T4]	
	AD13	T19]	
	AD17	U1]	
	AD21	U4]	
	AE7	U19]	
	AE8	U22]	
	AE19	W4]	
	AE20	W6]	
	AF11	W7		

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



	PIN NO.			eigna 2 coonputer (commuter,				
SIGNAL NAME	GJL/	GLS/	TYPE†	DESCRIPTION				
10 1112	GNZ	GNY						
SUPPLY VOLTAGE PINS (CONTINUED)								
	AF16	W9	S	3.3-V supply voltage (I/O)				
	-	W10						
	-	W13						
	-	W14						
	-	W16						
DV_DD	-	W17						
	-	W19						
	-	AB5						
	-	AB9						
	-	AB14						
	-	AB18						
	A1	E7						
	A2	E8						
	A3	E10						
	A24	E11						
	A25	E12						
	A26	E13						
	B1	E15						
	B2	E16						
	B3	F7						
	B24	F8						
	B25	F9						
	B26	F11						
	C1	F12						
	C2	F14		1.5. V gumbly voltage (core) (C6000P anh)				
CV _{DD}	C3	F15	S	1.5-V supply voltage (core) (C6202B only) 1.8-V supply voltage (core) (C6202 only)				
	C4	F16		, ,, ,,				
	C23	G5						
	C24	G6						
	C25	G17						
	C26	G18						
	D3	H5						
	D4	H6						
	D5	H17						
	D22	H18						
	D23	J6						
	D24	J17						
	E4	K5						
	E23	K18						
	AB4	L5						

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



0.0141	PIN NO.			, , , , , , , , , , , , , , , , , , , ,				
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE [†]	DESCRIPTION				
				SUPPLY VOLTAGE PINS (CONTINUED)				
	AB23	L6						
	AC3	L17						
	AC4	L18						
	AC5	M5						
	AC22	M6						
	AC23	M17						
	AC24	M18						
	AD1	N5						
	AD2	N18						
	AD3	P6						
	AD4	P17						
	AD23	R5						
	AD24	R6						
	AD25	R17]					
	AD26	R18						
	AE1	T5						
	AE2	T6		1.5 V gunghy voltage (core) (CS200P only)				
CV _{DD}	AE3	T17	S	1.5-V supply voltage (core) (C6202B only) 1.8-V supply voltage (core) (C6202 only)				
	AE24	T18						
	AE25	U7						
	AE26	U8						
	AF1	U9						
	AF2	U11						
	AF3	U12						
	AF24	U14						
	AF25	U15						
	AF26	U16						
	_	V7						
	_	V8						
	_	V10						
	_	V11						
		V12						
	_	V13						
	_	V15						
	-	V16						
	GROUND PINS							
	A4	A1	- GND					
V_{SS}	A8	A5		Ground pins				
	A13	A12						
	A14	A18						

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground



SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

PIN NO.				Signal Descriptions (Continued)				
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION				
GROUND PINS (CONTINUED)								
	A15	A22						
	A19	B2						
	A23	B21						
	B4	C1						
	B12	C3						
	B13	C20						
	B14	C22						
	B23	D5						
	C5	D8						
	C11 [‡]	D11						
	C16	D12						
	C22	D15						
	D1	D18						
	D2	E4						
	D6	E5						
	D21	E6						
	D25	E9						
V_{SS}	D26	E14	GND	Ground pins				
33	E3	E17		•				
	E24	E18						
	F4	E19						
	F23	F5						
	H1	F6						
	H26	F10	- - -					
	K1	F13						
	K26	F17						
	M1	F18						
	M26 N1	H4 H19						
	N2	J1						
	N25	J5						
	N26	J18						
	P1	J22						
	P2	K6						
	P25	K17						
	P26	L4						

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground ‡ For the C6202 GJL package only, the C11 pin is ground (V_{SS}). For all C62x™ GNZ packages, the C11 pin is CLKMODE1.

010	PIN	NO.		orginal Decompliants (Continues)				
SIGNAL NAME	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION				
GROUND PINS (CONTINUED)								
	R1	L19						
	R26	M4						
	U1	M19						
	U26	N6	-					
	W1	N17						
	W26	P1						
	AA4	P5						
	AA23	P18						
	AB3	P22						
	AB24	R4						
	AC1	R19						
	AC2	U5						
	AC6	U6						
	AC21	U10	GND					
	AC25	U13						
	AC26	U17						
	AD5	U18						
	AD22	V4						
V_{SS}	AE4	V5		Ground pins				
	AE13	V6						
	AE14	V9						
	AE23	V14						
	AF4	V17						
	AF8	V18						
	AF10	V19						
	AF12	W5						
	AF13	W8						
	AF14	W11						
	AF15	W12						
	AF17	W15						
	AF19	W18						
	AF23	Y1						
	_	Y3						
	_	Y20						
		Y22						
		AA2						
	_	AA21						

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground

SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

Signal Descriptions (Continued)

SIGNAL NAME	PIN NO.						
	GJL/ GNZ	GLS/ GNY	TYPE†	DESCRIPTION			
GROUND PINS (CONTINUED)							
	_	AB1		Ground pins			
	_	AB3					
	_	AB7	GND				
V _{SS}	_	AB11					
	_	AB16					
	_	AB20					
	_	AB22					

[†] I = Input, O = Output, Z = High Impedance, S = Supply Voltage, GND = Ground

development support

TI offers an extensive line of development tools for the TMS320C6000 DSP platform, including tools to evaluate the performance of the processors, generate code, develop algorithm implementations, and fully integrate and debug software and hardware modules.

The following products support development of C6000™ DSP-based applications:

Software Development Tools:

Code Composer Študio[™] Integrated Development Environment (IDE) including Editor C/C++/Assembly Code Generation, and Debug plus additional development tools Scalable, Real-Time Foundation Software (DSP/BIOS[™]), which provides the basic run-time target software needed to support any DSP application.

Hardware Development Tools:

Extended Development System (XDS™) Emulator (supports C6000 DSP multiprocessor system debug) EVM (Evaluation Module)

For a complete listing of development-support tools for the TMS320C6000 DSP platform, visit the Texas Instruments web site on the Worldwide Web at http://www.ti.com uniform resource locator (URL). For information on pricing and availability, contact the nearest TI field sales office or authorized distributor.

Code Composer Studio, DSP/BIOS, XDS, and TMS320 are trademarks of Texas Instruments.



device and development-support tool nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all TMS320 DSP devices and support tools. Each TMS320 DSP commercial family member has one of three prefixes: TMX, TMP, or TMS. Texas Instruments recommends two of three possible prefix designators for support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (TMX/TMDX) through fully qualified production devices/tools (TMS/TMDS).

Device development evolutionary flow:

TMX Experimental device that is not necessarily representative of the final device's electrical

specifications

TMP Final silicon die that conforms to the device's electrical specifications but has not completed

quality and reliability verification

TMS Fully qualified production device

Support tool development evolutionary flow:

TMDX Development-support product that has not yet completed Texas Instruments internal qualification

testing.

TMDS Fully qualified development-support product

TMX and TMP devices and TMDX development-support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

TMS devices and TMDS development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (TMX or TMP) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, GLS), the temperature range (for example, blank is the default commercial temperature range), and the device speed range in megahertz (for example, -250 is 250 MHz).

Table 16 lists the device orderable part numbers (P/Ns) and Figure 4 provides a legend for reading the complete device name for any member of the TMS320C6000 DSP platform. For more information on the C6202/02B device orderable P/Ns, visit the Texas Instruments web site on the Worldwide web at http://www.ti.com URL, or contact the nearest TI field sales office or authorized distributor.

device and development-support tool nomenclature (continued)

Table 16. TMS320C6202/02B Device Part Numbers (P/Ns) and Ordering Information

DEVICE ORDERABLE P/N	DEVICE SPEED	CV _{DD} (CORE VOLTAGE)	DV _{DD} (I/O VOLTAGE)	OPERATING CASE TEMPERATURE RANGE							
	C6202										
TMS320C6202GJL200	200 MHz/1600 MIPS	1.8 V	3.3 V	0°C to 90°C							
TMS320C6202GJL250	250 MHz/2000 MIPS	1.8 V	3.3 V	0°C to 90°C							
TMS320C6202GJLA200	200 MHz/1600 MIPS	1.8 V	3.3 V	−40°C to105°C							
TMS320C6202GLS200	200 MHz/1600 MIPS	1.8 V	3.3 V	0°C to 90°C							
TMS320C6202GLS250	250 MHz/2000 MIPS	1.8 V	3.3 V	0°C to 90°C							
	C6202B										
TMS320C6202BGNY250	250 MHz/2000 MIPS	1.5 V	3.3 V	0°C to 90°C							
TMS320C6202BGNY300	300 MHz/2400 MIPS	1.5 V	3.3 V	0°C to 90°C							
TMS320C6202BGNZ250	250 MHz/2000 MIPS	1.5 V	3.3 V	0°C to 90°C							
TMS320C6202BGNZ300	300 MHz/2400 MIPS	1.5 V	3.3 V	0°C to 90°C							
TMS32C6202BGNZA250	250 MHz/2000 MIPS	1.5 V	3.3 V	-40°C to 105°C							

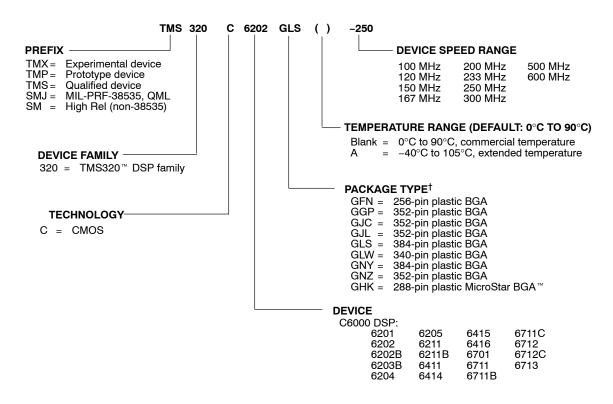


Figure 4. TMS320C6000™ DSP Platform Device Nomenclature (Including TMS320C6202 and C6202B)

MicroStar BGA is a trademark of Texas Instruments.

† BGA = Ball Grid Array



TMS320C6202, TMS320C6202B FIXED-POINT DIGITAL SIGNAL PROCESSORS

SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

documentation support

Extensive documentation supports all TMS320 DSP family devices from product announcement through applications development. The types of documentation available include: data sheets, such as this document, with design specifications; complete user's reference guides for all devices and tools; technical briefs; development-support tools; on-line help; and hardware and software applications. The following is a brief, descriptive list of support documentation specific to the C6000 DSP devices:

The *TMS320C6000 CPU and Instruction Set Reference Guide* (literature number SPRU189) describes the C6000 CPU (DSP core) architecture, instruction set, pipeline, and associated interrupts.

The *TMS320C6000 DSP Peripherals Overview Reference Guide* (literature number SPRU190) briefly describes the functionality of the peripherals available on the C6000 DSP platform of devices, such as the 64-/32-/16-bit external memory interfaces (EMIFs), 32-/16-bit host-port interfaces (HPIs), multichannel buffered serial ports (McBSPs), direct memory access (DMA), enhanced direct-memory-access (EDMA) controller, expansion bus (XBus), peripheral component interconnect (PCI), clocking and phase-locked loop (PLL); and power-down modes.

The How to Begin Development Today and Migrate Across the TMS320C6202/02B/03B/04 DSPs Application Report (literature number SPRA603) describes the migration concerns and identifies the similarities and differences between the C6202, C6202B, C6203B, and C6204 C6000 DSP devices.

The *TMS320C6202, TMS320C6202B Digital Signal Processors Silicon Errata* (literature number SPRZ152) describes the known exceptions to the functional specifications for particular silicon revisions of the TMS320C6202 device. There are currently *no* known silicon advisories on the TMS320C6202B device.

The *Using IBIS Models for Timing Analysis Application Report* (literature number SPRA839) describes how to properly use IBIS models to attain accurate timing analysis for a given system.

The tools support documentation is electronically available within the Code Composer Studio™ IDE. For a complete listing of the latest C6000 DSP documentation, visit the Texas Instruments web site on the Worldwide Web at http://www.ti.com uniform resource locator (URL).

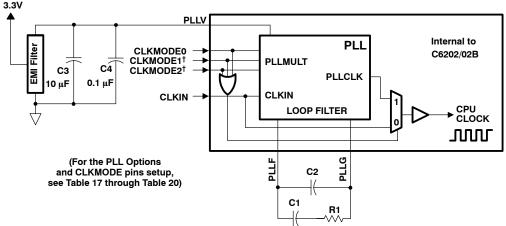
clock PLL

All of the internal C6202/02B clocks are generated from a single source through the CLKIN pin. This source clock either drives the PLL, which multiplies the source clock in frequency to generate the internal CPU clock, or bypasses the PLL to become the internal CPU clock.

To use the PLL to generate the CPU clock, the external PLL filter circuit must be properly designed. Figure 5, and Table 17 through Table 20 show the external PLL circuitry for either x1 (PLL bypass) or x4 PLL multiply modes. Figure 6 shows the external PLL circuitry for a system with ONLY x1 (PLL bypass) mode.

To minimize the clock jitter, a single clean power supply should power both the C6202/02B device and the external clock oscillator circuit. Noise coupling into PLLF directly impacts PLL clock jitter. The minimum CLKIN rise and fall times should also be observed. For the input clock timing requirements, see the Input and Output Clocks electricals section.

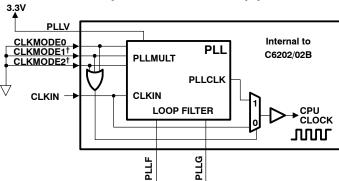
clock PLL (continued)



† CLKMODE1 and CLKMODE2 pins are not applicable (N/A) to the C6202 GJL package. The CLKMODE2 pin is also N/A on the C6202B GNZ package.

- NOTES: A. Keep the lead length and the number of vias between pin PLLF, pin PLLG, R1, C1, and C2 to a minimum. In addition, place all PLL components (R1, C1, C2, C3, C4, and EMI Filter) as close to the C6000 DSP device as possible. Best performance is achieved with the PLL components on a single side of the board without jumpers, switches, or components other than the ones shown.
 - B. For reduced PLL jitter, maximize the spacing between switching signals and the PLL external components (R1, C1, C2, C3, C4, and the EMI Filter).
 - C. The 3.3-V supply for the EMI filter must be from the same 3.3-V power plane supplying the I/O voltage, DV_{DD}.
 - D. EMI filter manufacturer: TDK part number ACF451832-333, 223, 153, 103. Panasonic part number EXCCET103U.

Figure 5. External PLL Circuitry for Either PLL Multiply Modes or x1 (Bypass) Mode



† CLKMODE1 and CLKMODE2 pins are not applicable (N/A) to the C6202 GJL package. The CLKMODE2 pin is also N/A on the C6202B GNZ package.

- NOTES: A. For a system with ONLY PLL x1 (bypass) mode, short the PLLF to PLLG.
 - B. The 3.3-V supply for PLLV must be from the same 3.3-V power plane supplying the I/O voltage, DV_{DD}.

Figure 6. External PLL Circuitry for x1 (Bypass) PLL Mode Only



clock PLL (continued)

Table 17. TMS320C6202 GLS and C6202B GNY Packages PLL Multiply and Bypass (x1) Options[†]

	GLS PACKAGE – 18 X 18 MM BGA [C6202 ONLY] AND GNY PACKAGE – 18 X 18 MM BGA [C6202B ONLY]								
DEVICES AND PLL									
BIT (PIN NO.)	CLKMODE2 (A14)	CLKMODE1 (A9)	CLKMODE0 (B12)	C6202 (GLS)‡	C6202B (GNY)				
	0	0	0	Bypass (x1)	Bypass (x1)				
	0	0	1	x4	x4				
	0	1	0	Bypass (x1)	x8				
	0	1	1	x4	x10				
Value	1	0	0	Bypass (x1)	х6				
	1	0	1	x4	х9				
	1	1	0	Bypass (x1)	x7				
	1	1	1	x4	x11				

[†] f(CPU Clock) = f(CLKIN) x (PLL mode)

Table 18. TMS320C6202 GJL and TMS320C6202B GNZ Packages PLL Multiply and Bypass (x1) Options[†]

GJL PACKAGE 27 x 27 mm BGA [C6202 only] and GNZ PACKAGE – 27 x 27 mm BGA [C6202B only]									
DEVICES AND PLL CLOCK OPTIONS									
BIT (PIN NO.)	CLKMODE2 (N/A)§	E2 (N/A) [§] CLKMODE1 (C11) [§] CLKMODE0 (B1		C6202 (GJL)§	C6202B (GNZ)§				
		0	0	Bypass (x1)	Bypass (x1)				
Value	NI/A	0	1	x4	x4				
value	N/A	1	0	NI/A	x8				
		1	1	N/A	x10				

[†] f(CPU Clock) = f(CLKIN) x (PLL mode)

Table 19. TMS320C6202 PLL Component Selection Table[†]

CLKMODE	CLKIN RANGE (MHz)	CPU CLOCK FREQUENCY (CLKOUT1) RANGE (MHz)	CLKOUT2 RANGE (MHz)	R1 [±1%]	C1 [±10%]	C2 [±10%]	TYPICAL LOCK TIME (μs)
x4	32.5-62.5	130-250	65-125	60.4 Ω	27 nF	560 pF	75

[†] Under some operating conditions, the maximum PLL lock time may vary by as much as 150% from the specified typical value. For example, if the typical lock time is specified as 100 μs, the maximum value may be as long as 250 μs.

For the C6202 GLS package, the CLKMODE2 (A14) and CLKMODE1 (A9) pins are internally unconnected.

[§] CLKMODE2 and CLKMODE1 pins are not available on the C6202 GJL package.

The CLKMODE2 pin is not available (N/A) on the C6202B GNZ package.

clock PLL (continued)

Table 20. TMS320C6202B PLL Component Selection Table[†]

CLKMODE‡	CLKIN RANGE (MHz)	CPU CLOCK FREQUENCY RANGE (MHz)	CLKOUT2 RANGE (MHz)	R1 [±1%]	C1 [±10%]	C2 [±10%]	TYPICAL LOCK TIME (μs)
x4	32.5–75						
x6	21.7–50						
x7	18.6-42.9						
x8	16.3–37.5	130–300	65-150	45.3 Ω	47 nF	10 pF	75
х9	14.4–33.3						
x10	13–30						
x11	11.8–27.3						

[†] Under some operating conditions, the maximum PLL lock time may vary by as much as 150% from the specified typical value. For example, if the typical lock time is specified as 100 μs, the maximum value may be as long as 250 μs.

power-down mode logic

Figure 7 shows the power-down logic on for the 6202/6202B.

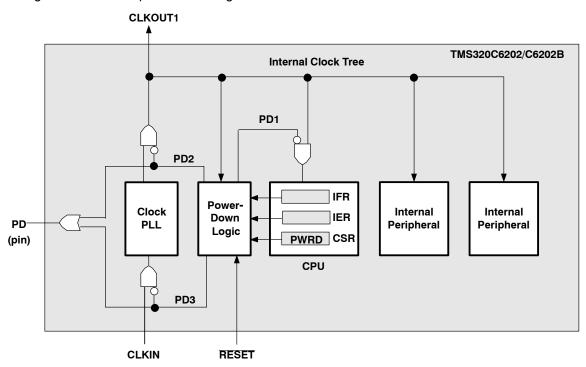
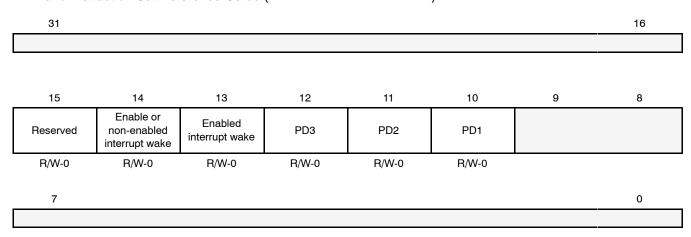


Figure 7. Power-Down Mode Logic

[‡] CLKMODE x1, x4, x6, x7, x8, x9, x10, and x11 apply to the C6202B GNY devices. The C6202B GNZ device is restricted to x1, x4, x8, and x10 multiply factors.

triggering, wake-up, and effects

The power-down modes and their wake-up methods are programmed by setting the PWRD field (bits 10–15) of the control status register (CSR). The PWRD field of the CSR is shown in Figure 8 and described in Table 21. When writing to the CSR, all bits of the PWRD field should be set at the same time. Logic 0 should be used when writing to the reserved bit (bit 15) of the PWRD field. The CSR is discussed in detail in the *TMS320C6000 CPU* and *Instruction Set Reference Guide* (literature number SPRU189).



Legend: R/W-x = Read/write reset value

NOTE: The shadowed bits are not part of the power-down logic discussion and therefore are not covered here. For information on these other bit fields in the CSR register, see the *TMS320C6000 CPU and Instruction Set Reference Guide* (literature number SPRU189).

Figure 8. PWRD Field of the CSR Register

Power-down mode PD1 takes effect eight to nine clock cycles after the instruction that sets the PWRD bits in the CSR.

If PD1 mode is terminated by a non-enabled interrupt, the program execution returns to the instruction where PD1 took effect. If PD1 mode is terminated by an enabled interrupt, the interrupt service routine will be executed first, then the program execution returns to the instruction where PD1 took effect. The GIE bit in CSR and the NMIE bit in the interrupt enable register (IER) must also be set in order for the interrupt service routine to execute; otherwise, execution returns to the instruction where PD1 took effect upon PD1 mode termination by an enabled interrupt.

PD2 and PD3 modes can only be aborted by device reset. Table 21 summarizes all the power-down modes.

triggering, wake-up, and effects (continued)

Table 21. Characteristics of the Power-Down Modes

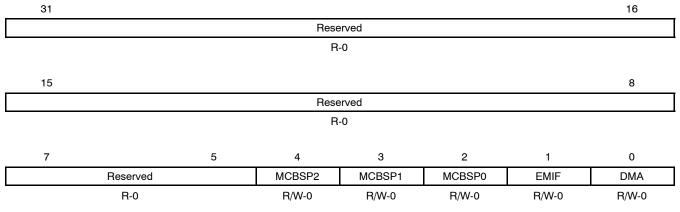
PRWD BITS 15-10	POWER-DOWN MODE	WAKE-UP METHOD	EFFECT ON CHIP'S OPERATION
000000	No power-down	_	
001001	PD1	Wake by an enabled interrupt	CPU halted (except for the interrupt logic) Power-down mode blocks the internal clock inputs at the
010001	PD1	Wake by an enabled or non-enabled interrupt	boundary of the CPU, preventing most of its logic from switching. During PD1, DMA transactions can proceed between peripherals and internal memory.
011010	PD2 [†]	Wake by a device reset	Output clock from PLL is halted, stopping the internal clock structure from switching and resulting in the entire chip being halted. Signal terminal PD is driven high. All register and internal RAM contents are preserved. All functional I/O "freeze" in the last state when the PLL clock is turned off. ‡
011100	PD3 [†]	Wake by a device reset	Input clock to the PLL stops generating clocks. Signal terminal PD is driven high. All register and internal RAM contents are preserved. All functional I/O "freeze" in the last state when the PLL clock is turned off. [†] Following reset, the PLL needs time to re-lock, just as it does following power-up. Wake-up from PD3 takes longer than wake-up from PD2 because the PLL needs to be re-locked.
other	Reserved	_	

[†] On the C6202/C6202B, both the PD2 and PD3 signals assert the PD pin for external recognition of these two power-down modes.

peripheral power-down mode for TMS320C6202/6202B

The C6202/C6202B has the ability to turn off clocks to individual peripherals on the device. This feature allows the user to selectively turn off peripherals which are not being used for a specific application and not pay the extra price in power consumption for unused peripherals.

The Figure 9 title displays the peripheral power down register address location and Figure 9 itself shows the register fields.



Legend: R/W-x = Read/write reset value

Figure 9. Peripheral Power-Down Control Register (PDCTL) for TMS320C6202/6202B (019C 0200h)



[‡] When entering PD2 and PD3, all functional I/O will remain in the previous state. However, for peripherals which are asynchronous in nature (HPI) or peripherals with an external clock source (McBSP, XBUS, timers), output signals may transition in response to stimulus on the inputs. Peripheral operation may not perform as intended under these conditions.

peripheral power-down mode for TMS320C6202/6202B (continued)

Table 22 lists and describes the fields in the peripheral power-down control register (PDCTL).

Table 22. Power-Down Control Register (PDCTL) Field Descriptions

BIT	FIELD	VALUES	DESCRIPTION
31-5	Reserved		Reserved. The reserved bit location is always read as zero. A value written to this field has no effect.
4	MCBSP2		Internal McBSP2 clock enable bit.
		0	Internal McBSP2 clock is enabled.
		1	Internal McBSP2 clock is disabled. McBSP2 is not functional.
3	MCBSP1		Internal McBSP1 clock enable bit.
		0	Internal McBSP1 clock is enabled.
		1	Internal McBSP1 clock is disabled. McBSP1 is not functional.
2	MCBSP0		Internal McBSP0 clock enable bit.
		0	Internal McBSP0 clock is enabled.
		1	Internal McBSP0 clock is disabled. McBSP1 is not functional.
1	EMIF		Internal EMIF clock enable bit.
		0	Internal EMIF clock is enabled.
		1	Internal EMIF clock is disabled. EMIF is not functional.
0	DMA		Internal DMA clock enable bit.
		0	Internal DMA clock is enabled.
		1	Internal DMA clock is disabled. DMA is not functional.

The user must be careful to not disable a portion of the device which is being used, since the peripheral becomes non-operational once disabled. A clock-off mode can be entered and exited depending on the needs of the application. For example, if an application does not need the serial ports, the ports can be disabled and then re-enabled when needed. While a peripheral is in power-down mode, no writes to the peripheral's registers will occur; and reads from the peripheral will produce invalid data.

When re-enabling any of the peripheral power-down bits, the CPU should wait at least 5 additional clock cycles before attempting to access that peripheral.

power-supply sequencing

TI DSPs do not require specific power sequencing between the core supply and the I/O supply. However, systems should be designed to ensure that neither supply is powered up for extended periods of time (>1 second) if the other supply is below the proper operating voltage.

system-level design considerations

System-level design considerations, such as bus contention, may require supply sequencing to be implemented. In this case, the core supply should be powered up at the same time as, or prior to (and powered down after), the I/O buffers. This is to ensure that the I/O buffers receive valid inputs from the core before the output buffers are powered up, thus, preventing bus contention with other chips on the board.

power-supply design considerations

A dual-power supply with simultaneous sequencing can be used to eliminate the delay between core and I/O power up. A Schottky diode can also be used to tie the core rail to the I/O rail (see Figure 10).

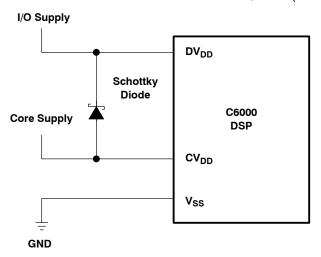


Figure 10. Schottky Diode Diagram

Core and I/O supply voltage regulators should be located close to the DSP (or DSP array) to minimize inductance and resistance in the power delivery path. Additionally, when designing for high-performance applications utilizing the C6000 platform of DSPs, the PC board should include separate power planes for core, I/O, and ground, all bypassed with high-quality low-ESL/ESR capacitors.

On systems using C62x and C67x DSPs, the core may consume in excess of 2 A per DSP until the I/O supply powers on. This extra current results from uninitialized logic within the DSP(s). A normal current state returns once the I/O power supply turns on and the CPU sees a clock pulse. Decreasing the amount of time between the core supply power-up and the I/O supply power-up reduces the effects of the current draw. If the external supply to the DSP core cannot supply the excess current, the minimum core voltage may not be achieved until after normal current returns. This voltage starvation of the core supply during power up does not affect run-time operation. Voltage starvation can affect power supply systems that gate the I/O supply via the core supply, causing the I/O supply to never turn on. During the transition from excess to normal current, a voltage spike may be seen on the core supply. Care must be taken when designing overvoltage protection circuitry on the core supply to not restart the power sequence due to this spike. Otherwise, the supply may cycle indefinitely.

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SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

IEEE 1149.1 JTAG compatibility statement

For compatibility with IEEE 1149.1 JTAG programmers, the \overline{TRST} pin may need to be externally pulled up via a 1-k Ω resistor. For these C62x devices, this pin is internally pulled down, holding the JTAG port in reset by default. This is typically only a problem in systems where the DSP shares a scan chain with some other device. Some JTAG programmers for these other devices do not actively drive \overline{TRST} high, leaving the scan chain inoperable while the C62x JTAG port is held in reset. TI emulators do drive \overline{TRST} high, so the external pullup resistor is not needed in systems where TI emulators are the only devices that control JTAG scan chains on which the DSP(s) reside. If the system has other devices in the same scan chain as the DSP, and the programmer for these devices does *not* drive \overline{TRST} high, then an external 1-k Ω pullup resistor is required.

With this external 1-k Ω pullup resistor installed, care must be taken to keep the DSP in a usable state under all circumstances. When \overline{TRST} is pulled up, the JTAG driver must maintain the TMS signal high for 5 TCLK cycles, forcing the DSP(s) into the test logic reset (TLR) state. From the TLR state, the DSP's data scan path can be put in bypass (scan all 1s into the IR) to scan the other devices. The TLR state also allows normal operation of the DSP. If operation without anything driving the JTAG port is desired, the pullup resistor should be jumpered so that it can be engaged for programming the other devices and disconneted for running without a JTAG programmer or emulator.

TMS320C6202, TMS320C6202B FIXED-POINT DIGITAL SIGNAL PROCESSORS

SPRS104I - OCTOBER 1999 - REVISED MARCH 2004

absolute maximum ratings over operating case temperature ranges (unless otherwise noted)[†]

Supply voltage range, CV _{DD} (see Note 1)	– 0.3 V to 2.3 V
Supply voltage range, DV _{DD} (see Note 1)	–0.3 V to 4 V
Input voltage range	–0.3 V to 4 V
Output voltage range	–0.3 V to 4 V
Operating case temperature ranges, T _C : (default)	0°C to 90°C
(A version): C6202BGNZA-250	–40°C to105°C
Storage temperature range, T _{stq}	65°C to 150°C
Temperature cycle range, (1000-cycle performance)	40°C to 125°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to VSS.

recommended operating conditions

			MIN	NOM	MAX	UNIT
CV _{DD}	Supply voltage, Core	C6202B only	1.43	1.5	1.57	V
CV _{DD}	Supply voltage, Core	C6202 only	1.71	1.8	1.89	V
DV_DD	Supply voltage, I/O		3.14	3.3	3.46	V
V_{SS}	Supply ground		0	0	0	V
V _{IH}	High-level input voltage		2			V
V _{IL}	Low-level input voltage				0.8	V
I _{OH}	High-level output current				-8	mA
I _{OL}	Low-level output current				8	mA
_	Occupies and bound to	Default	0		90	°C
T _C	Operating case temperature	A version: C6202BGNZA-250	-40		105	°C

electrical characteristics over recommended ranges of supply voltage and operating case temperature (unless otherwise noted)

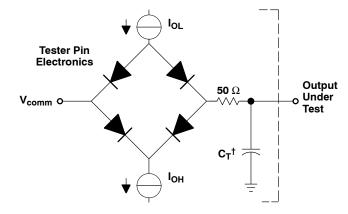
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	$DV_{DD} = MIN, I_{OH} = MAX$	2.4			V
V_{OL}	Low-level output voltage	$DV_{DD} = MIN,$ $I_{OL} = MAX$			0.6	V
II	Input current [‡]	$V_I = V_{SS}$ to DV_{DD}			±10	uA
l _{OZ}	Off-state output current	$V_O = DV_{DD}$ or 0 V			±10	uA
	Supply current, CPU + CPU memory	C6202, CV _{DD} = NOM, CPU clock = 200 MHz		±10 520 340 390		mA
I _{DD2V}	access§	C6202B, CV _{DD} = NOM, CPU clock = 200 MHz		340		mA
	0	C6202, CV _{DD} = NOM, CPU clock = 200 MHz		390		mA
I _{DD2V}	Supply current, peripherals§	C6202B, CV _{DD} = NOM, CPU clock = 200 MHz		235		mA
	0 1 1/0 \$	C6202, DV _{DD} = NOM, CPU clock = 200 MHz		70		mA
IDD3V	Supply current, I/O pins§	C6202B, DV _{DD} = NOM, CPU clock = 200 MHz		45		mA
C _i	Input capacitance				10	pF
Co	Output capacitance				10	pF

[‡] TMS and TDI are not included due to internal pullups. TRST is not included due to internal pulldown.



[§] Measured with average activity (50% high / 50% low power). For more details on CPU, peripheral, and I/O activity, see the *TMS320C62x/C67x Power Consumption Summary* application report (literature number SPRA486).

PARAMETER MEASUREMENT INFORMATION



Where: I_{OL} = 2 mA I_{OH} = 2 mA V_{comm} = 1.5 V

C_T = 15-pF typical load-circuit capacitance

Figure 11. Test Load Circuit for AC Timing Measurements

signal transition levels

All input and output timing parameters are referenced to 1.5 V for both "0" and "1" logic levels.

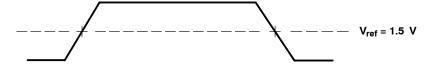


Figure 12. Input and Output Voltage Reference Levels for ac Timing Measurements

All rise and fall transition timing parameters are referenced to V_{IL} MAX and V_{IH} MIN for input clocks, and V_{OL} MAX and V_{OH} MIN for output clocks.

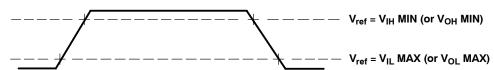


Figure 13. Rise and Fall Transition Time Voltage Reference Levels

[†] Typical distributed load circuit capacitance

PARAMETER MEASUREMENT INFORMATION (CONTINUED)

timing parameters and board routing analysis

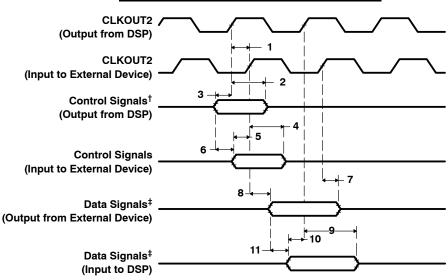
The timing parameter values specified in this data sheet do *not* include delays by board routings. As a good board design practice, such delays must *always* be taken into account. Timing values may be adjusted by increasing/decreasing such delays. TI recommends utilizing the available I/O buffer information specification (IBIS) models to analyze the timing characteristics correctly. To properly use IBIS models to attain accurate timing analysis for a given system, see the *Using IBIS Models for Timing Analysis Application Report* (literature number SPRA839). If needed, external logic hardware such as buffers may be used to compensate any timing differences.

For inputs, timing is most impacted by the round-trip propagation delay from the DSP to the external device and from the external device to the DSP. This round-trip delay tends to negatively impact the input setup time margin, but also tends to improve the input hold time margins (see Table 23 and Figure 14).

Figure 14 represents a general transfer between the DSP and an external device. The figure also represents board route delays and how they are perceived by the DSP and the external device.

DESCRIPTION Clock route delay 1 2 Minimum DSP hold time 3 Minimum DSP setup time 4 External device hold time requirement External device setup time requirement 5 6 Control signal route delay 7 External device hold time 8 External device access time 9 DSP hold time requirement 10 DSP setup time requirement Data route delay 11

Table 23. Board-Level Timings Example (see Figure 14)



[†] Control signals include data for Writes.

Figure 14. Board-Level Input/Output Timings



 $[\]ensuremath{\ddagger}$ Data signals are generated during Reads from an external device.

INPUT AND OUTPUT CLOCKS

timing requirements for CLKIN (PLL used)^{†‡§} (see Figure 15)

NO.				2-200	C6202-250 C6202B-250		C6202B-300		UNIT
			MIN	MAX	MIN	MAX	MIN	MAX	
1	t _{c(CLKIN)}	Cycle time, CLKIN	5 * M		4 * M		3.33 * M		ns
2	t _{w(CLKINH)}	Pulse duration, CLKIN high	0.4C		0.4C		0.4C		ns
3	t _{w(CLKINL)}	Pulse duration, CLKIN low	0.4C		0.4C		0.4C		ns
4	t _{t(CLKIN)}	Transition time, CLKIN		5		5		5	ns

 $[\]dagger$ The reference points for the rise and fall transitions are measured at V_{IL} MAX and V_{IH} MIN.

timing requirements for CLKIN [PLL bypassed (x1)]^{\dagger ¶} (see Figure 15)

NO.			C6202	2-200	C6202		C6202I	B-300	UNIT
			MIN	MAX	MIN	MAX	MIN	MAX	
1	t _{c(CLKIN)}	Cycle time, CLKIN	5		4		3.33		ns
2	t _{w(CLKINH)}	Pulse duration, CLKIN high	0.45C		0.45C		0.45C		ns
3	t _{w(CLKINL)}	Pulse duration, CLKIN low	0.45C		0.45C		0.45C		ns
4	t _{t(CLKIN)}	Transition time, CLKIN		0.6		0.6		0.6	ns

[†] The reference points for the rise and fall transitions are measured at V_{II} MAX and V_{IH} MIN.

C = CLKIN cycle time in ns. For example, when CLKIN frequency is 50 MHz, use C = 20 ns. The maximum CLKIN cycle time is PLL bypass mode (x1) is 200 MHz.

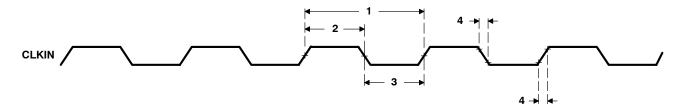


Figure 15. CLKIN Timings

[‡] M = the PLL multiplier factor (x4, x6, x7, x8, x9, x10, or x11) for C6202 GLS and C6202B GNY only.

M = the PLL multiplier factor (x4, x6, x8, or x10) for C6202B GNZ only.

For more details, see the Clock PLL section of this data sheet.

[§] C = CLKIN cycle time in ns. For example, when CLKIN frequency is 50 MHz, use C = 20 ns.

INPUT AND OUTPUT CLOCKS (CONTINUED)

timing requirements for XCLKIN[†] (see Figure 16)

NO.			C6202 C6202 C6202 C6202	2-250 B-250	UNIT
			MIN	MAX	
1	t _{c(XCLKIN)}	Cycle time, XCLKIN	4P		ns
2	t _{w(XCLKINH)}	Pulse duration, XCLKIN high	1.8P		ns
3	t _{w(XCLKINL)}	Pulse duration, XCLKIN low	1.8P		ns

[†] P = 1/CPU clock frequency in nanoseconds (ns).

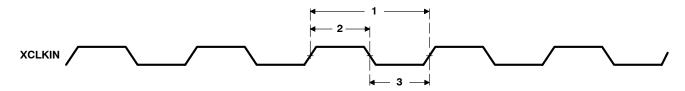


Figure 16. XCLKIN Timings

switching characteristics over recommended operating conditions for CLKOUT2^{‡§} (see Figure 17)

NO.	NO. PARAMETER		C6202- C6202- C6202B C6202B	UNIT	
				MAX	
2	t _{w(CKO2H)}	Pulse duration, CLKOUT2 high	P – 0.7	P + 0.7	ns
3	t _{w(CKO2L)}	Pulse duration, CLKOUT2 low	P – 0.7	P + 0.7	ns

P = 1/CPU clock frequency in ns.

 $[\]S$ The reference points for the rise and fall transitions are measured at V_{OL} MAX and V_{OH} MIN.

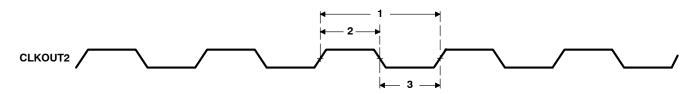


Figure 17. CLKOUT2 Timings

INPUT AND OUTPUT CLOCKS (CONTINUED)

switching characteristics over recommended operating conditions for XFCLK^{†‡} (see Figure 18)

NO.	IO. PARAMETER		C620 C620 C6202 C6202	UNIT	
			MIN	MAX	
1	t _{c(XFCK)}	Cycle time, XFCLK	D*P-0.7	D*P+0.7	ns
2	t _{w(XFCKH)}	Pulse duration, XFCLK high	(D/2) * P - 0.7	(D/2) * P + 0.7	ns
3	t _{w(XFCKL)}	Pulse duration, XFCLK low	(D/2) * P - 0.7	(D/2) * P + 0.7	ns

[†] P = 1/CPU clock frequency in ns.

[‡] D = 8, 6, 4, or 2; FIFO clock divide ratio, user-programmable

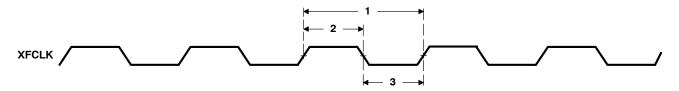


Figure 18. XFCLK Timings

ASYNCHRONOUS MEMORY TIMING

timing requirements for asynchronous memory cycles^{†‡§¶} (see Figure 19 – Figure 22)

NO.			C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT
			MIN	MAX	MIN	MAX	
3	t _{su(EDV-AREH)}	Setup time, EDx valid before ARE high	1		1		ns
4	t _{h(AREH-EDV)}	Hold time, EDx valid after ARE high	3.5		4.9		ns
6	t _{su(ARDYH-AREL)}	Setup time, ARDY high before ARE low	-[(RST - 3) * P - 6]		-[(RST - 3) * P - 6]		ns
7	t _{h(AREL-ARDYH)}	Hold time, ARDY high after ARE low	(RST – 3) * P + 2		(RST – 3) * P + 2		ns
9	t _{su(ARDYL-AREL)}	Setup time, ARDY low before ARE low	-[(RST - 3) * P - 6]		-[(RST - 3) * P - 6]		ns
10	t _{h(AREL-ARDYL)}	Hold time, ARDY low after ARE low	(RST – 3) * P + 2		(RST – 3) * P + 2		ns
11	t _{w(ARDYH)}	Pulse width, ARDY high	2P		2P		ns
15	t _{su(ARDYH-AWEL)}	Setup time, ARDY high before AWE low	-[(WST - 3) * P - 6]		-[(WST - 3) * P - 6]		ns
16	t _{h(AWEL-ARDYH)}	Hold time, ARDY high after AWE low	(WST - 3) * P + 2		(WST - 3) * P + 2		ns
18	t _{su(ARDYL-AWEL)}	Setup time, ARDY low before AWE low	-[(WST - 3) * P - 6]	_	-[(WST - 3) * P - 6]	_	ns
19	t _{h(AWEL-ARDYL)}	Hold time, ARDY low after AWE low	(WST - 3) * P + 2		(WST - 3) * P + 2		ns

[†] To ensure data setup time, simply program the strobe width wide enough. ARDY is internally synchronized. If ARDY does meet setup or hold time, it may be recognized in the current cycle or the next cycle. Thus, ARDY can be an asynchronous input.

[‡] RS = Read Setup, RST = Read Strobe, RH = Read Hold, WS = Write Setup, WST = Write Strobe, WH = Write Hold. These parameters are programmed via the EMIF CE space control registers.

[§] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

The sum of RS and RST (or WS and WST) must be a minimum of 4 in order to use ARDY input to extend strobe width.

ASYNCHRONOUS MEMORY TIMING (CONTINUED)

switching characteristics over recommended operating conditions for asynchronous memory cycles^{†‡§¶} (see Figure 19 – Figure 22)

NO.		PARAMETER	C6202-200 C6202-250		C6202B-250 C6202B-300			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX	
1	t _{osu(SELV-AREL)}	Output setup time, select signals valid to ARE low	RS * P - 2			RS * P - 2			ns
2	t _{oh(AREH-SELIV)}	Output hold time, ARE high to select signals invalid	RH * P - 2			RH*P-2			ns
5	t _{w(AREL)}	Pulse width, ARE low		RST * P			RST * P		ns
8	t _{d(ARDYH-AREH)}	Delay time, ARDY high to ARE high	3P		4P + 5	3P		4P + 5	ns
12	t _{osu(SELV-AWEL)}	Output setup time, select signals valid to \overline{AWE} low	WS * P - 3			WS * P - 2			ns
13	t _{oh(AWEH-SELIV)}	Output hold time, AWE high to select signals invalid	WH * P – 2			WH * P - 2			ns
14	t _{w(AWEL)}	Pulse width, AWE low		WST*P			WST*P		ns
17	t _{d(ARDYH-AWEH)}	Delay time, ARDY high to AWE high	3P		4P + 5	3P		4P + 5	ns

[†] RS = Read Setup, RST = Read Strobe, RH = Read Hold, WS = Write Setup, WST = Write Strobe, WH = Write Hold. These parameters are programmed via the EMIF CE space control registers.

 $^{^{\}ddagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[§] The sum of RS and RST (or WS and WST) must be a minimum of 4 in order to use ARDY input to extend strobe width.

Select signals include: $\overline{\text{CEx}}$, $\overline{\text{BE}[3:0]}$, EA[21:2], $\overline{\text{AOE}}$; and for writes, include ED[31:0], with the exception that $\overline{\text{CEx}}$ can stay active for an additional 7P ns following the end of the cycle.

ASYNCHRONOUS MEMORY TIMING (CONTINUED)

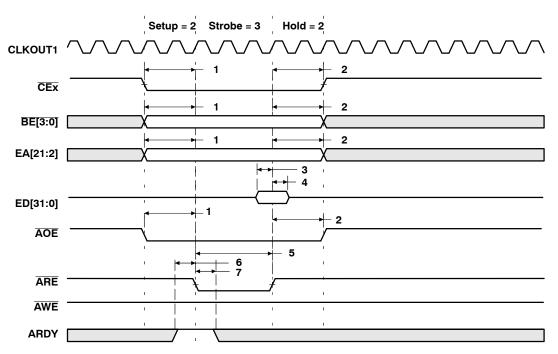


Figure 19. Asynchronous Memory Read Timing (ARDY Not Used)

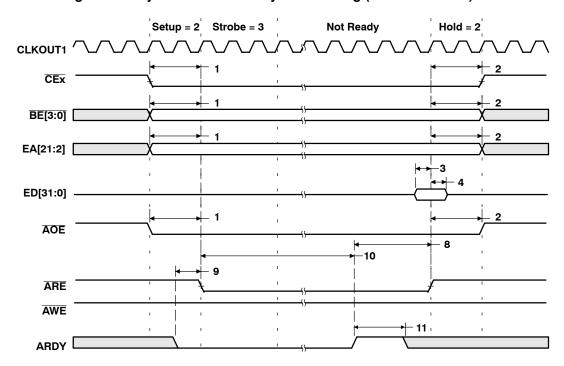


Figure 20. Asynchronous Memory Read Timing (ARDY Used)

ASYNCHRONOUS MEMORY TIMING (CONTINUED)

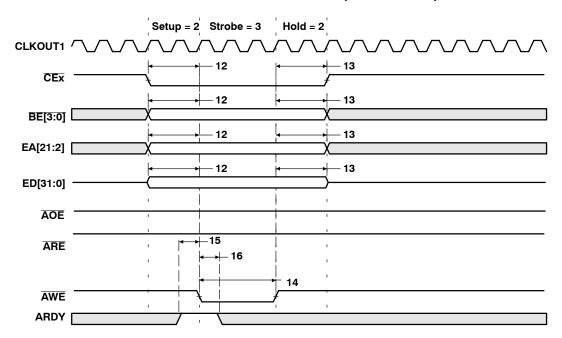


Figure 21. Asynchronous Memory Write Timing (ARDY Not Used)

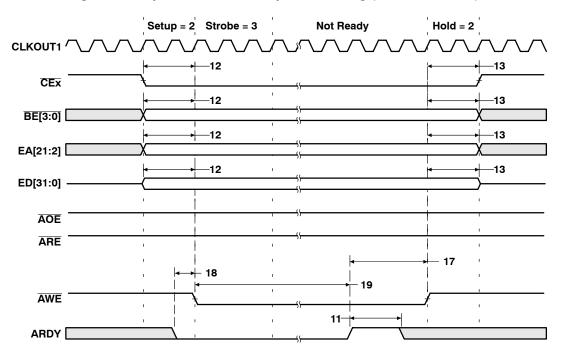


Figure 22. Asynchronous Memory Write Timing (ARDY Used)

SYNCHRONOUS-BURST MEMORY TIMING

timing requirements for synchronous-burst SRAM cycles for C6202 devices (see Figure 23)

NO.	NO.		C6202	2-200	C6202-250		
NO.			MIN	MAX	MIN	MAX	UNIT
7	t _{su(EDV-CKO2H)}	Setup time, read EDx valid before CLKOUT2 high	2.5		2.0		ns
8	t _{h(CKO2H-EDV)}	Hold time, read EDx valid after CLKOUT2 high	2.0		2.0		ns

switching characteristics over recommended operating conditions for synchronous-burst SRAM cycles for C6202 devices^{†‡} (see Figure 23 and Figure 24)

		24244	C6202	-200	C6202-	-250	
NO.		PARAMETER	MIN	MAX	MIN	MAX	UNIT
1	t _{osu(CEV-CKO2H)}	Output setup time, CEx valid before CLKOUT2 high	P – 0.8		P – 0.8		ns
2	toh(CKO2H-CEV)	Output hold time, CEx valid after CLKOUT2 high	P – 4		P – 3		ns
3	t _{osu(BEV-CKO2H)}	Output setup time, BEx valid before CLKOUT2 high	P – 0.8		P – 0.8		ns
4	t _{oh(CKO2H-BEIV)}	Output hold time, BEx invalid after CLKOUT2 high	P – 4		P – 3		ns
5	t _{osu(EAV-CKO2H)}	Output setup time, EAx valid before CLKOUT2 high	P – 0.8		P – 0.8		ns
6	t _{oh(CKO2H-EAIV)}	Output hold time, EAx invalid after CLKOUT2 high	P – 4		P – 3		ns
9	t _{osu(ADSV-CKO2H)}	Output setup time, SDCAS/SSADS valid before CLKOUT2 high	P – 0.8		P – 0.8		ns
10	t _{oh(CKO2H-ADSV)}	Output hold time, SDCAS/SSADS valid after CLKOUT2 high	P – 4		P – 3		ns
11	t _{osu(OEV-CKO2H)}	Output setup time, SDRAS/SSOE valid before CLKOUT2 high	P – 0.8		P – 0.8		ns
12	t _{oh(CKO2H-OEV)}	Output hold time, SDRAS/SSOE valid after CLKOUT2 high	P – 4		P – 3		ns
13	t _{osu(EDV-CKO2H)}	Output setup time, EDx valid before CLKOUT2 high§	P – 1.2		P – 1.2		ns
14	t _{oh(CKO2H-EDIV)}	Output hold time, EDx invalid after CLKOUT2 high	P – 4		P – 3		ns
15	t _{osu(WEV-CKO2H)}	Output setup time, SDWE/SSWE valid before CLKOUT2 high	P - 0.8		P – 0.8		ns
16	t _{oh(CKO2H-WEV)}	Output hold time, SDWE/SSWE valid after CLKOUT2 high	P – 4		P – 3		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[‡] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SSADS, SSOE, and SSWE, respectively, during SBSRAM accesses.

[§] For the first write in a series of one or more consecutive adjacent writes, the write data is generated one CLKOUT2 cycle early to accommodate the ED enable time.

SYNCHRONOUS-BURST MEMORY TIMING (CONTINUED)

timing requirements for synchronous-burst SRAM cycles C6202B devices (see Figure 23)

NO			6202I	3-250	C6202B-300		
NO.		ı	MIN	MAX	MIN	MAX	UNIT
7	t _{su(EDV-CKO2H)} Setup time, read EDx valid before CLKOUT2 high		2.9		1.6		ns
8	t _{h(CKO2H-EDV)} Hold time, read EDx valid after CLKOUT2 high		2.1		2.3		ns

switching characteristics over recommended operating conditions for synchronous-burst SRAM cycles for C6202B devices^{†‡} (see Figure 23 and Figure 24)

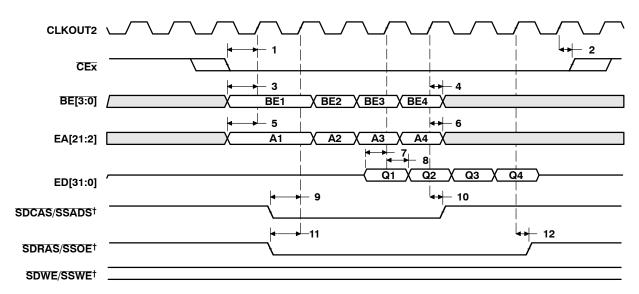
		DADAMETER	C6202E	3-250	C6202B	3-300	
NO.		PARAMETER	MIN	MAX	MIN	MAX	UNIT
1	t _{osu(CEV-CKO2H)}	Output setup time, CEx valid before CLKOUT2 high	P – 1.7		P – 1		ns
2	t _{oh(CKO2H-CEV)}	Output hold time, CEx valid after CLKOUT2 high	P – 3.4		P – 2.7		ns
3	t _{osu(BEV-CKO2H)}	Output setup time, BEx valid before CLKOUT2 high	P – 1.7		P – 1		ns
4	t _{oh(CKO2H-BEIV)}	Output hold time, BEx invalid after CLKOUT2 high	P – 3.4		P – 2.7		ns
5	t _{osu(EAV-CKO2H)}	Output setup time, EAx valid before CLKOUT2 high	P – 1.7		P – 1		ns
6	t _{oh(CKO2H-EAIV)}	Output hold time, EAx invalid after CLKOUT2 high	P – 3.4		P – 2.7		ns
9	t _{osu(ADSV-CKO2H)}	Output setup time, SDCAS/SSADS valid before CLKOUT2 high	P – 1.7		P – 1		ns
10	t _{oh(CKO2H-ADSV)}	Output hold time, SDCAS/SSADS valid after CLKOUT2 high	P – 3.4		P – 2.7		ns
11	t _{osu(OEV-CKO2H)}	Output setup time, SDRAS/SSOE valid before CLKOUT2 high	P – 1.7		P – 1		ns
12	t _{oh(CKO2H-OEV)}	Output hold time, SDRAS/SSOE valid after CLKOUT2 high	P – 3.4		P – 2.7		ns
13	t _{osu(EDV-CKO2H)}	Output setup time, EDx valid before CLKOUT2 high§	P – 2.3		P – 1.6		ns
14	toh(CKO2H-EDIV)	Output hold time, EDx invalid after CLKOUT2 high	P – 3.2		P – 2.5		ns
15	t _{osu(WEV-CKO2H)}	Output setup time, SDWE/SSWE valid before CLKOUT2 high	P – 1.7	_	P – 1	_	ns
16	t _{oh(CKO2H-WEV)}	Output hold time, SDWE/SSWE valid after CLKOUT2 high	P – 3.4		P – 2.7		ns

 $^{^\}dagger$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[‡] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SSADS, SSOE, and SSWE, respectively, during SBSRAM accesses.

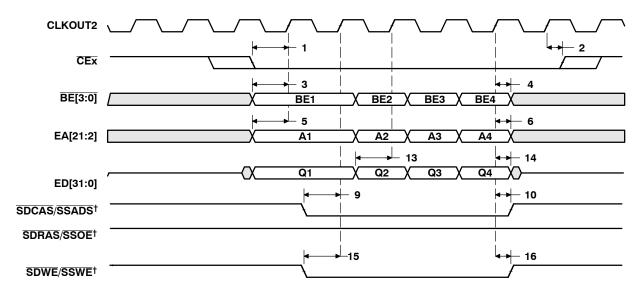
[§] For the first write in a series of one or more consecutive adjacent writes, the write data is generated one CLKOUT2 cycle early to accommodate the ED enable time.

SYNCHRONOUS-BURST MEMORY TIMING (CONTINUED)



[†] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SSADS, SSOE, and SSWE, respectively, during SBSRAM accesses.

Figure 23. SBSRAM Read Timing



 $^{^{\}dagger} \overline{\text{SDCAS}/\text{SSADS}}, \overline{\text{SDRAS}/\text{SSOE}}, \text{ and } \overline{\text{SDWE}/\text{SSWE}} \text{ operate as } \overline{\text{SSADS}}, \overline{\text{SSOE}}, \text{ and } \overline{\text{SSWE}}, \text{ respectively, during SBSRAM accesses.}$

Figure 24. SBSRAM Write Timing

SYNCHRONOUS DRAM TIMING

timing requirements for synchronous DRAM cycles for C6202 devices (see Figure 25)

	NO		C6202-250	
NO.		MIN MA	K MIN MAX	UNIT
7	t _{su(EDV-CKO2H)} Setup time, read EDx valid before CLKOUT2 high	1.2	1.2	ns
8	t _{h(CKO2H-EDV)} Hold time, read EDx valid after CLKOUT2 high	3	2.7	ns

switching characteristics over recommended operating conditions for synchronous DRAM cycles for C6202 devices^{†‡} (see Figure 25–Figure 30)

	DADAMETER	C6202-	-200	C6202-250		
	PARAMETER	MIN	MAX	MIN	MAX	UNIT
t _{osu(CEV-CKO2H)}	Output setup time, CEx valid before CLKOUT2 high	P – 1		P – 0.9		ns
t _{oh(CKO2H-CEV)}	Output hold time, CEx valid after CLKOUT2 high	P – 3.5		P – 2.9		ns
t _{osu(BEV-CKO2H)}	Output setup time, BEx valid before CLKOUT2 high	P – 1		P – 0.9		ns
toh(CKO2H-BEIV)	Output hold time, BEx invalid after CLKOUT2 high	P – 3.5		P – 2.9		ns
t _{osu(EAV-CKO2H)}	Output setup time, EAx valid before CLKOUT2 high	P – 1		P – 0.9		ns
toh(CKO2H-EAIV)	Output hold time, EAx invalid after CLKOUT2 high	P – 3.5		P – 2.9		ns
t _{osu(CASV-CKO2H)}	Output setup time, SDCAS/SSADS valid before CLKOUT2 high	P – 1		P - 0.9		ns
toh(CKO2H-CASV)	Output hold time, SDCAS/SSADS valid after CLKOUT2 high	P – 3.5		P – 2.9		ns
t _{osu(EDV-CKO2H)}	Output setup time, EDx valid before CLKOUT2 high§	P – 1		P – 1.5		ns
toh(CKO2H-EDIV)	Output hold time, EDx invalid after CLKOUT2 high	P – 3.5		P – 2.8		ns
t _{osu(WEV-CKO2H)}	Output setup time, SDWE/SSWE valid before CLKOUT2 high	P – 1		P - 0.9		ns
t _{oh(CKO2H-WEV)}	Output hold time, SDWE/SSWE valid after CLKOUT2 high	P – 3.5		P – 2.9		ns
t _{osu(SDA10V-CKO2H)}	Output setup time, SDA10 valid before CLKOUT2 high	P – 1		P – 0.9		ns
toh(CKO2H-SDA10IV)	Output hold time, SDA10 invalid after CLKOUT2 high	P – 3.5		P – 2.9		ns
t _{osu(RASV-CKO2H)}	Output setup time, SDRAS/SSOE valid before CLKOUT2 high	P – 1		P - 0.9		ns
t _{oh(CKO2H-RASV)}	Output hold time, SDRAS/SSOE valid after CLKOUT2 high	P – 3.5		P – 2.9		ns
	toh(CKO2H-CEV) tosu(BEV-CKO2H) toh(CKO2H-BEIV) toh(CKO2H-BEIV) toh(CKO2H-EAIV) toh(CKO2H-EAIV) toh(CKO2H-CASV) toh(CKO2H-CASV) tosu(EDV-CKO2H) toh(CKO2H-EDIV) toh(CKO2H-EDIV) tosu(WEV-CKO2H) toh(CKO2H-WEV) tosu(SDA10V-CKO2H) toh(CKO2H-SDA10IV) tosu(RASV-CKO2H)	toh(CKO2H-CEV) Output hold time, CEx valid after CLKOUT2 high tosu(BEV-CKO2H) Output setup time, BEx invalid after CLKOUT2 high toh(CKO2H-BEIV) Output setup time, EAx valid before CLKOUT2 high tosu(EAV-CKO2H) Output hold time, EAx invalid after CLKOUT2 high toh(CKO2H-EAIV) Output setup time, EAx invalid after CLKOUT2 high Output setup time, SDCAS/SSADS valid before CLKOUT2 high Output hold time, SDCAS/SSADS valid after CLKOUT2 high toh(CKO2H-CASV) Output hold time, EDx valid before CLKOUT2 high toh(CKO2H-EDIV) Output setup time, EDx invalid after CLKOUT2 high Output setup time, SDWE/SSWE valid before CLKOUT2 high Output setup time, SDWE/SSWE valid after CLKOUT2 high Output hold time, SDWE/SSWE valid after CLKOUT2 high Output hold time, SDWE/SSWE valid after CLKOUT2 high toh(CKO2H-WEV) Output setup time, SDWE/SSWE valid after CLKOUT2 high Output setup time, SDWE/SSWE valid after CLKOUT2 high Output setup time, SDA10 valid before CLKOUT2 high toh(CKO2H-SDA10IV) Output setup time, SDA10 invalid after CLKOUT2 high Output setup time, SDRAS/SSOE valid before CLKOUT2 high Output hold time, SDRAS/SSOE valid after CLKOUT2	Tosu(CEV-CKO2H) Tosu(CEV-CKO2H) Output setup time, CEx valid before CLKOUT2 high P - 1 Toh(CKO2H-CEV) Output hold time, EEx valid after CLKOUT2 high P - 3.5 Tosu(BEV-CKO2H) Output setup time, BEx valid before CLKOUT2 high P - 1 Toh(CKO2H-BEIV) Output hold time, BEx invalid after CLKOUT2 high P - 3.5 Tosu(EAV-CKO2H) Output setup time, EAx valid before CLKOUT2 high P - 1 Toh(CKO2H-EAIV) Output hold time, EAx invalid after CLKOUT2 high P - 3.5 Tosu(CASV-CKO2H) Output setup time, SDCAS/SSADS valid before CLKOUT2 high Toh(CKO2H-CASV) Toutput hold time, SDCAS/SSADS valid after CLKOUT2 Toki(CKO2H-CASV) Output setup time, EDx valid before CLKOUT2 high P - 3.5 Tosu(EDV-CKO2H) Output setup time, EDx valid before CLKOUT2 high P - 3.5 Tosu(WEV-CKO2H) Output hold time, EDx invalid after CLKOUT2 high P - 3.5 Tosu(WEV-CKO2H) Output hold time, SDWE/SSWE valid before CLKOUT2 Toh(CKO2H-WEV) Output hold time, SDWE/SSWE valid after CLKOUT2 Toh(CKO2H-WEV) Output setup time, SDWE/SSWE valid after CLKOUT2 Toh(CKO2H-WEV) Output setup time, SDA10 valid before CLKOUT2 high P - 3.5 Tosu(RASV-CKO2H) Output setup time, SDA10 invalid after CLKOUT2 high P - 3.5 Tosu(RASV-CKO2H) Output setup time, SDRAS/SSOE valid before CLKOUT2 high Output hold time, SDRAS/SSOE valid after CLKOUT2 P - 3.5	tosu(CEV-CKO2H) Output setup time, CEx valid before CLKOUT2 high P - 1 toh(CKO2H-CEV) Output hold time, CEx valid after CLKOUT2 high P - 3.5 tosu(BEV-CKO2H) Output setup time, BEx valid before CLKOUT2 high P - 3.5 tosu(EAV-CKO2H) Output setup time, BEx invalid after CLKOUT2 high P - 3.5 tosu(EAV-CKO2H) Output setup time, EAx valid before CLKOUT2 high P - 3.5 tosu(CASV-CKO2H) Output hold time, EAx invalid after CLKOUT2 high P - 3.5 tosu(CASV-CKO2H) Output hold time, SDCAS/SSADS valid before CLKOUT2 high P - 1 toh(CKO2H-CASV) high Output hold time, SDCAS/SSADS valid after CLKOUT2 high P - 3.5 tosu(EDV-CKO2H) Output setup time, EDx valid before CLKOUT2 high P - 3.5 tosu(EDV-CKO2H) Output setup time, EDx invalid after CLKOUT2 high P - 3.5 tosu(WEV-CKO2H) Output setup time, SDWE/SSWE valid before CLKOUT2 high Output hold time, SDWE/SSWE valid after CLKOUT2 high P - 3.5 tosu(SDA10V-CKO2H) Output setup time, SDWE/SSWE valid after CLKOUT2 high Output setup time, SDWE/SSWE valid before CLKOUT2 high P - 3.5 tosu(SDA10V-CKO2H) Output setup time, SDA10 valid before CLKOUT2 high P - 3.5 tosu(RASV-CKO2H) Output setup time, SDRAS/SSOE valid after CLKOUT2 high Output setup time, SDRAS/SSOE valid before CLKOUT2 high Output setup time, SDRAS/SSOE valid after CLKOUT2 P - 3.5	PARAMETER MIN MAX MIN tosu(CEV-CK02H) Output setup time, CEx valid before CLKOUT2 high P - 1 toh(CK02H-CEV) Output hold time, CEx valid after CLKOUT2 high P - 3.5 P - 2.9 tosu(BEV-CK02H) Output setup time, BEx invalid after CLKOUT2 high P - 3.5 P - 2.9 toh(CK02H-BEIV) Output setup time, BEx invalid after CLKOUT2 high P - 3.5 P - 2.9 tosu(EAV-CK02H) Output setup time, EAx valid before CLKOUT2 high P - 3.5 P - 2.9 toh(CK02H-EAIV) Output hold time, EAx invalid after CLKOUT2 high P - 3.5 P - 2.9 toh(CK02H-EAIV) Output hold time, SDCAS/SSADS valid before CLKOUT2 high P - 1 P - 0.9 toh(CK02H-CASV) Output hold time, SDCAS/SSADS valid after CLKOUT2 high tosu(EDV-CK02H) Output hold time, EDx valid before CLKOUT2 high P - 3.5 P - 2.9 tosu(WEV-CK02H) Output hold time, EDx invalid after CLKOUT2 high P - 3.5 P - 2.8 toh(CK02H-WEV) Output setup time, SDWE/SSWE valid before CLKOUT2 high Coutput hold time, SDWE/SSWE valid after CLKOUT2 high P - 3.5 P - 2.9 toh(CK02H-WEV) Output hold time, SDWE/SSWE valid after CLKOUT2 high P - 3.5 P - 2.9 toh(CK02H-SDA10IV) Output setup time, SDA10 invalid after CLKOUT2 high P - 3.5 P - 2.9 toh(CK02H-SDA10IV) Output setup time, SDRAS/SSOE valid before CLKOUT2 high P - 3.5 P - 2.9	Tesu(CEV-CKO2H) Output setup time, CEx valid before CLKOUT2 high P-1 P-0.9 toh(CKO2H-CEV) Output hold time, CEx valid after CLKOUT2 high P-1 P-0.9 tosu(BEV-CKO2H) Output setup time, BEx valid before CLKOUT2 high P-1 P-0.9 toh(CKO2H-BEIV) Output hold time, BEx invalid after CLKOUT2 high P-1 P-0.9 toh(CKO2H-BEIV) Output setup time, EAx valid before CLKOUT2 high P-1 P-0.9 toh(CKO2H-BAIV) Output hold time, EAx invalid after CLKOUT2 high P-1 P-0.9 toh(CKO2H-BAIV) Output setup time, SDCAS/SSADS valid before CLKOUT2 high P-1 P-0.9 toh(CKO2H-CASV) Output hold time, SDCAS/SSADS valid after CLKOUT2 high toh(CKO2H-CASV) Output setup time, EDx valid before CLKOUT2 high P-1 P-1.5 toh(CKO2H-CRO2H) Output setup time, EDx valid before CLKOUT2 high P-3.5 P-2.9 tosu(EDV-CKO2H) Output setup time, EDx invalid after CLKOUT2 high P-3.5 P-2.8 tosu(WEV-CKO2H) Output setup time, SDWE/SSWE valid after CLKOUT2 high Output hold time, SDWE/SSWE valid after CLKOUT2 high Toh(CKO2H-WEV) Output hold time, SDWE/SSWE valid after CLKOUT2 high Output setup time, SDWE/SSWE valid after CLKOUT2 high Toh(CKO2H-SDA10IV) Output setup time, SDA10 valid before CLKOUT2 high P-1 P-0.9 tosu(SDA10V-CKO2H) Output setup time, SDA10 valid before CLKOUT2 high P-1 P-0.9 toh(CKO2H-SDA10IV) Output setup time, SDA10 valid before CLKOUT2 high Output setup time, SDA3/SSOE valid before CLKOUT2 high Output hold time, SDA3/SSOE valid after CLKOUT2 P-1 P-0.9

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[‡] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

[§] For the first write in a series of one or more consecutive adjacent writes, the write data is generated one CLKOUT2 cycle early to accommodate the ED enable time.

SYNCHRONOUS DRAM TIMING (CONTINUED)

timing requirements for synchronous DRAM cycles for C6202B devices (see Figure 25)

		C	C6202B-250		C6202E	LINUT	
NO.			MIN	MAX	MIN	MAX	UNIT
7	$t_{su(\text{EDV-CKO2H})}$ Setup time, read EDx valid before CLKOUT2 high		1.3		0		ns
8	t _{h(CKO2H-EDV)} Hold time, read EDx valid after CLKOUT2 high		2.3		2.3		ns

switching characteristics over recommended operating conditions for synchronous DRAM cycles for C6202B devices^{†‡} (see Figure 25–Figure 30)

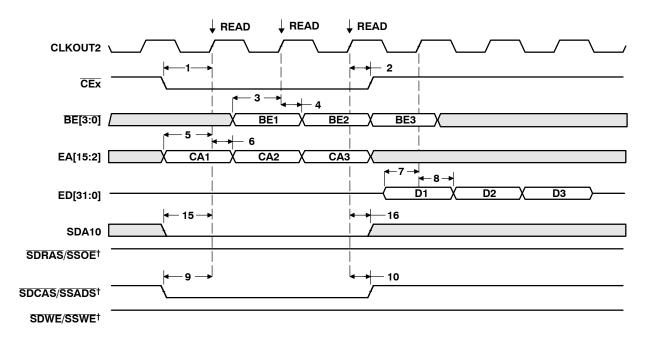
		DADAMETER	C6202E	3-250	C6202E	3-300	UNIT
NO.		PARAMETER	MIN	MAX	MIN	MAX	UNII
1	t _{osu(CEV-CKO2H)}	Output setup time, CEx valid before CLKOUT2 high	P – 1.7		P – 1		ns
2	toh(CKO2H-CEV)	Output hold time, CEx valid after CLKOUT2 high	P – 3		P – 2.3		ns
3	t _{osu(BEV-CKO2H)}	Output setup time, BEx valid before CLKOUT2 high	P – 1.7		P – 1		ns
4	toh(CKO2H-BEIV)	Output hold time, BEx invalid after CLKOUT2 high	P – 3		P – 2.3		ns
5	t _{osu(EAV-CKO2H)}	Output setup time, EAx valid before CLKOUT2 high	P – 1.7		P – 1		ns
6	t _{oh(CKO2H-EAIV)}	Output hold time, EAx invalid after CLKOUT2 high	P – 3		P – 2.3		ns
9	t _{osu(CASV-CKO2H)}	Output setup time, SDCAS/SSADS valid before CLKOUT2 high	P – 1.7		P – 1		ns
10	toh(CKO2H-CASV)	Output hold time, SDCAS/SSADS valid after CLKOUT2 high	P – 3		P – 2.3		ns
11	t _{osu(EDV-CKO2H)}	Output setup time, EDx valid before CLKOUT2 high§	P – 2.3		P – 1.6		ns
12	toh(CKO2H-EDIV)	Output hold time, EDx invalid after CLKOUT2 high	P – 2.7		P – 2		ns
13	t _{osu(WEV-CKO2H)}	Output setup time, SDWE/SSWE valid before CLKOUT2 high	P – 1.7		P – 1		ns
14	toh(CKO2H-WEV)	Output hold time, SDWE/SSWE valid after CLKOUT2 high	P – 3		P – 2.3		ns
15	t _{osu(SDA10V-CKO2H)}	Output setup time, SDA10 valid before CLKOUT2 high	P – 1.7		P – 1		ns
16	toh(CKO2H-SDA10IV)	Output hold time, SDA10 invalid after CLKOUT2 high	P – 3		P – 2.3		ns
17	t _{osu(RASV-CKO2H)}	Output setup time, SDRAS/SSOE valid before CLKOUT2 high	P – 1.7		P – 1		ns
18	t _{oh(CKO2H-RASV)}	Output hold time, SDRAS/SSOE valid after CLKOUT2 high	P – 3		P – 2.3		ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

^{*} SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

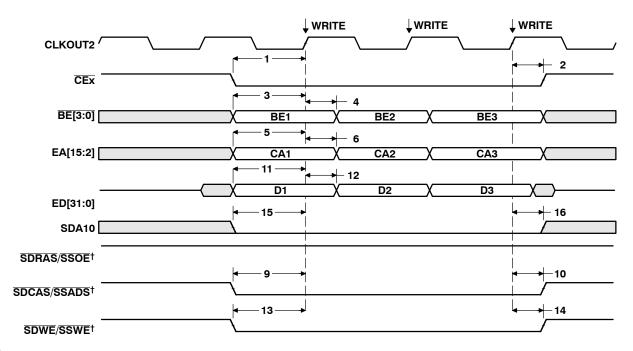
[§] For the first write in a series of one or more consecutive adjacent writes, the write data is generated one CLKOUT2 cycle early to accommodate the ED enable time.

SYNCHRONOUS DRAM TIMING (CONTINUED)



[†] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

Figure 25. Three SDRAM READ Commands

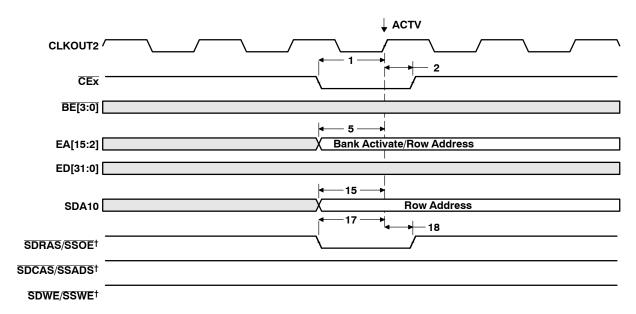


[†] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

Figure 26. Three SDRAM WRT Commands

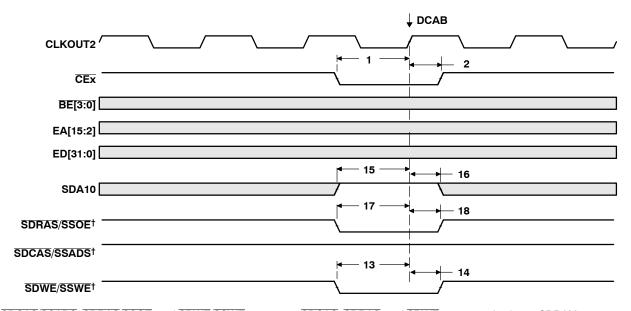


SYNCHRONOUS DRAM TIMING (CONTINUED)



[†] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

Figure 27. SDRAM ACTV Command

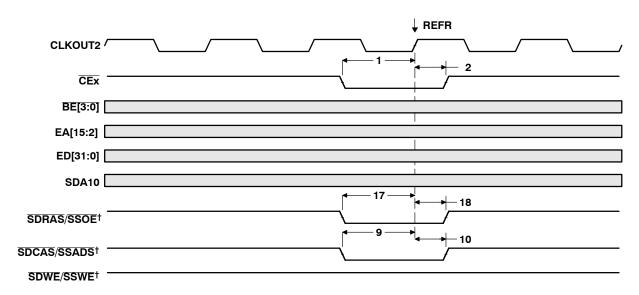


 $^{^{\}dagger} \overline{\text{SDCAS}}/\overline{\text{SSADS}}, \overline{\text{SDRAS}}/\overline{\text{SSOE}}, \text{ and } \overline{\text{SDWE}}/\overline{\text{SSWE}} \text{ operate as } \overline{\text{SDCAS}}, \overline{\text{SDRAS}}, \text{ and } \overline{\text{SDWE}}, \text{ respectively, during SDRAM accesses.}$

Figure 28. SDRAM DCAB Command

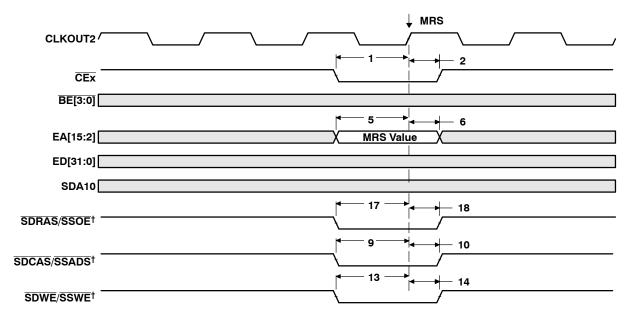


SYNCHRONOUS DRAM TIMING (CONTINUED)



[†] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

Figure 29. SDRAM REFR Command



[†] SDCAS/SSADS, SDRAS/SSOE, and SDWE/SSWE operate as SDCAS, SDRAS, and SDWE, respectively, during SDRAM accesses.

Figure 30. SDRAM MRS Command

HOLD/HOLDA TIMING

timing requirements for the HOLD/HOLDA cycles[†] (see Figure 31)

NO.	NO.		C6202-200 C6202-250 C6202B-250 C6202B-300		
		MIN	MAX		
3	$t_{oh(HOLDAL-HOLDL)}$ Output hold time, \overline{HOLD} low after \overline{HOLDA} low	Р		ns	

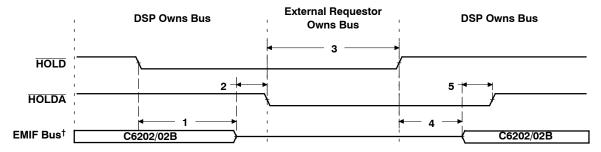
 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for the HOLD/HOLDA cycles^{†‡} (see Figure 31)

NO.	PARAMETER		C6202 C6202 C6202 C6202	UNIT	
			MIN	MAX	
1	t _{d(HOLDL-EMHZ)}	Delay time, HOLD low to EMIF Bus high impedance	3P	§	ns
2	t _d (EMHZ-HOLDAL)	Delay time, EMIF Bus high impedance to HOLDA low	0	2P	ns
4	t _{d(HOLDH-EMLZ)}	Delay time, HOLD high to EMIF Bus low impedance	3P	7P	ns
5	t _d (EMLZ-HOLDAH)	Delay time, EMIF Bus low impedance to HOLDA high	0	2P	ns

 $[\]frac{1}{P} = 1/CPU$ clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[§] All pending EMIF transactions are allowed to complete before HOLDA is asserted. The worst case for this is an asynchronous read or write with external ARDY used or a minimum of eight consecutive SDRAM reads or writes when RBTR8 = 1. If no bus transactions are occurring, then the minimum delay time can be achieved. Also, bus hold can be indefinitely delayed by setting NOHOLD = 1.



 $^{^{\}dagger} \text{ EMIF Bus consists of } \overline{\text{CE}[3:0]}, \overline{\text{BE}[3:0]}, \overline{\text{ED}[31:0]}, \overline{\text{EA}[21:2]}, \overline{\text{ARE}}, \overline{\text{AOE}}, \overline{\text{AWE}}, \overline{\text{SDCAS/SSADS}}, \overline{\text{SDRAS/SSOE}}, \overline{\text{SDWE/SSWE}}, \text{and SDA10}.$

Figure 31. HOLD/HOLDA Timing

[‡] EMIF Bus consists of CE[3:0], BE[3:0], ED[31:0], EA[21:2], ARE, AOE, AWE, SDCAS/SSADS, SDRAS/SSOE, SDWE/SSWE, and SDA10.

RESET TIMING

timing requirements for reset[†] (see Figure 32)

NO.			C6202-200 C6202-250 C6202B-250 C6202B-300		UNIT
		Width of the RESET pulse (PLL stable) [‡]	10P	WAX	ns
1	t _{w(RST)}	Width of the RESET pulse (PLL needs to sync up)§	250		μs
10	t _{su(XD)}	Setup time, XD configuration bits valid before RESET high [¶]	5P		ns
11	t _{h(XD)}	Hold time, XD configuration bits valid after RESET high [¶]	5P		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions during reset^{†#} (see Figure 32)

NO.	PANAMETEN		C6202 C6202 C6202 C6202	UNIT	
			MIN	MAX	
2	t _{d(RSTL-CKO2IV)}	Delay time, RESET low to CLKOUT2 invalid	Р		ns
3	t _{d(RSTH-CKO2V)}	Delay time, RESET high to CLKOUT2 valid		4P	ns
4	t _{d(RSTL-HIGHIV)}	Delay time, RESET low to high group invalid	Р		ns
5	t _{d(RSTH-HIGHV)}	Delay time, RESET high to high group valid		4P	ns
6	t _{d(RSTL-LOWIV)}	Delay time, RESET low to low group invalid	Р		ns
7	t _{d(RSTH-LOWV)}	Delay time, RESET high to low group valid		4P	ns
8	t _{d(RSTL-ZHZ)}	Delay time, RESET low to Z group high impedance	Р		ns
9	t _{d(RSTH-ZV)}	Delay time, RESET high to Z group valid		4P	ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

Low group consists of: IACK, INUM[3:0], DMAC[3:0], PD, TOUT0, and TOUT1

Z group consists of: EA[21:2], ED[31:0], CE[3:0], BE[3:0], ARE, AWE, AOE, SDCAS/SSADS, SDRAS/SSOE, SDWE/SSWE, SDA10, CLKX0, CLKX1, CLKX2, FSX0, FSX1, FSX2, DX0, DX1, DX2, CLKR0, CLKR1, CLKR2, FSR0, FSR1,

SDA10, CLKX0, CLKX1, CLKX2, FSX0, FSX1, FSX2, DX0, DX1, DX2, CLKR0, CLKR1, CLKR2, FSR0, FSR1, FSR2, XCE[3:0], XBE[3:0]/XA[5:2], XOE, XRE, XWE/XWAIT, XAS, XW/R, XRDY, XBLAST, XHOLD,

and XHOLDA



[‡] This parameter applies to CLKMODE x1 when CLKIN is stable, and applies to CLKMODE x4 only when CLKIN and PLL are stable for C6202. This parameter applies to CLKMODE x4, x6, x7, x8, x9, x10, and x11 when CLKIN and PLL are stable for C6202B GNY devices.

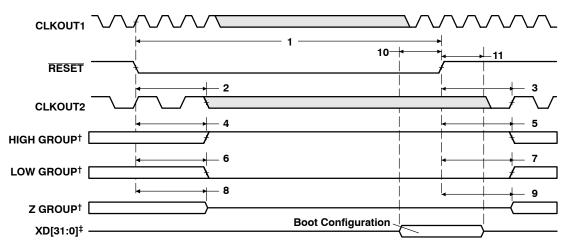
This parameter applies to CLKMODE x4, x6, x8, and x10 only when CLKIN and PLL are stable for C6202B GNZ devices.

[§] This parameter applies to CLKMODE x4 only (it does not apply to CLKMODE x1) for C6202. This parameter applies to CLKMODE x4, x6, x7, x8, x9, x10, and x11 only (it does not apply to CLKMODE x1) for C6202B GNY devices. This parameter applies to CLKMODE x4, x6, x8, and x10 only (it does not apply to CLKMODE x1) for C6202B GNZ devices. The RESET signal is not connected internally to the clock PLL circuit. The PLL, however, may need up to 250 μs to stabilize following device power up or after PLL configuration has been changed. During that time, RESET must be asserted to ensure proper device operation. See the Clock PLL section for PLL lock times.

 $^{^{\}P}$ XD[31:0] are the boot configuration pins during device reset.

[#] High group consists of: XFCLK, HOLDA

RESET TIMING (CONTINUED)



[†] High group consists of: Low group consists of:

 $\mathsf{XFCLK}, \overline{\mathsf{HOLDA}}$

IACK, INUM[3:0], DMAC[3:0], PD, TOUT0, and TOUT1.

Z group consists of:

EA[21:2], ED[31:0], CE[3:0], BE[3:0], ARE, AWE, AOE, SDCAS/SSADS, SDRAS/SSOE, SDWE/SSWE, SDA10, CLKX0, CLKX1, CLKX2, FSX0, FSX1, FSX2, DX0, DX1, DX2, CLKR0, CLKR1, CLKR2, FSR0, FSR1, FSR2, XCE[3:0], XBE[3:0]/XA[5:2], XOE, XRE, XWE/XWAIT, XAS, XW/R, XRDY, XBLAST, XHOLD, and XHOLDA.

Figure 32. Reset Timing



[‡] XD[31:0] are the boot configuration pins during device reset.

EXTERNAL INTERRUPT TIMING

timing requirements for interrupt response cycles[†] (see Figure 33)

NO.	NO.		UNIT
		MIN MAX	
2	t _{w(ILOW)} Width of the interrupt pulse low	2P	ns
3	t _{w(IHIGH)} Width of the interrupt pulse high	2P	ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions during interrupt response cycles[†] (see Figure 33)

NO.	PARAMETER		C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT
			MIN	MAX	MIN	MAX	
1	t _{R(EINTH - IACKH)}	Response time, EXT_INTx high to IACK high	9P		9P		ns
4	t _d (CKO2L-IACKV)	Delay time, CLKOUT2 low to IACK valid	1	10	-1.5	10	ns
5	t _d (CKO2L-INUMV)	Delay time, CLKOUT2 low to INUMx valid	0	10	-2.0	10	ns
6	t _d (CKO2L-INUMIV)	Delay time, CLKOUT2 low to INUMx invalid	0	10	-2.0	10	ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

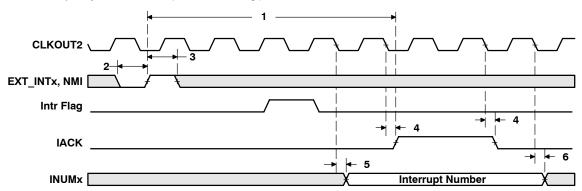


Figure 33. Interrupt Timing

EXPANSION BUS SYNCHRONOUS FIFO TIMING

timing requirements for synchronous FIFO interface (see Figure 34, Figure 35, and Figure 36)

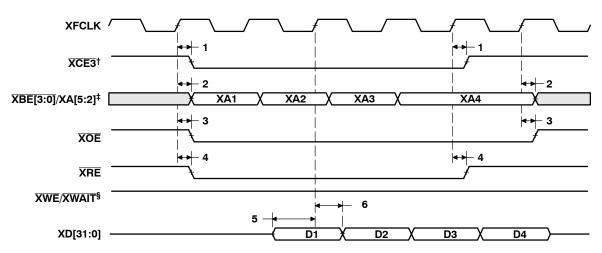
NO.			C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT
			MIN	MAX	MIN	MAX	
5	t _{su(XDV-XFCKH)}	Setup time, read XDx valid before XFCLK high	3		3		ns
6	t _{h(XFCKH-XDV)}	Hold time, read XDx valid after XFCLK high	2.5		2.5		ns

switching characteristics over recommended operating conditions for synchronous FIFO interface (see Figure 34, Figure 35, and Figure 36)

NO.	PARAMETER			C6202-200 C6202-250		C6202B-250 C6202B-300	
				MAX	MIN	MAX	
1	t _{d(XFCKH-XCEV)}	Delay time, XFCLK high to XCEx valid	1.5	5.2	1.5	5.5	ns
2	t _{d(XFCKH-XAV)}	Delay time, XFCLK high to XBE[3:0]/XA[5:2] valid [†]	1.5	5.2	1.5	5.5	ns
3	t _{d(XFCKH-XOEV)}	Delay time, XFCLK high to $\overline{\text{XOE}}$ valid	1.5	5.2	1.5	5.5	ns
4	t _{d(XFCKH-XREV)}	Delay time, XFCLK high to XRE valid	1.5	5.2	1.5	5.5	ns
7	t _{d(XFCKH-XWEV)}	Delay time, XFCLK high to XWE/XWAIT‡ valid	1.5	5.2	1.5	5.5	ns
8	t _{d(XFCKH-XDV)}	Delay time, XFCLK high to XDx valid		5.2		6	ns
9	t _{d(XFCKH-XDIV)}	Delay time, XFCLK high to XDx invalid	1.5		1.5		ns

[†] XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during synchronous FIFO accesses.

[‡] XWE/XWAIT operates as the write-enable signal XWE during synchronous FIFO accesses.



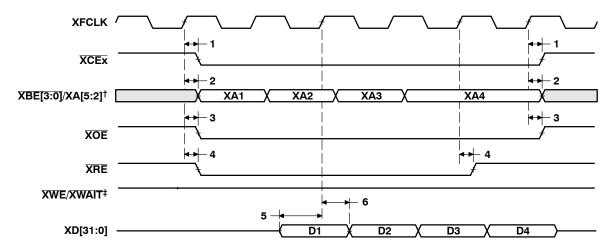
[†] FIFO read (glueless) mode only available in XCE3.

Figure 34. FIFO Read Timing (Glueless Read Mode)

[‡] XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during synchronous FIFO accesses.

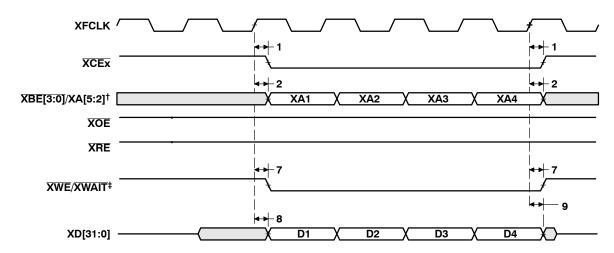
 $[\]S$ XWE/XWAIT operates as the write-enable signal XWE during synchronous FIFO accesses.

EXPANSION BUS SYNCHRONOUS FIFO TIMING (CONTINUED)



[†] XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during synchronous FIFO accesses.

Figure 35. FIFO Read Timing



 $^{^{\}dagger}$ $\overline{\text{XBE}[3:0]}/\text{XA}[5:2]$ operate as address signals XA[5:2] during synchronous FIFO accesses.

Figure 36. FIFO Write Timing

[‡] XWE/XWAIT operates as the write-enable signal XWE during synchronous FIFO accesses.

[‡] XWE/XWAIT operates as the write-enable signal XWE during synchronous FIFO accesses.

SPRS104I - OCTOBER 1999 - REVISED FEBRUARY 2004

EXPANSION BUS ASYNCHRONOUS PERIPHERAL TIMING

timing requirements for asynchronous peripheral cycles^{†‡§¶} (see Figure 37–Figure 40)

NO.			C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT
			MIN	MAX	MIN	MAX	
3	t _{su(XDV-XREH)}	Setup time, XDx valid before $\overline{\text{XRE}}$ high	4.5		4.5		ns
4	t _{h(XREH-XDV)}	Hold time, XDx valid after XRE high	1		2.5		ns
6	t _{su(XRDYH-XREL)}	Setup time, XRDY high before XRE low	-[(RST - 3) * P - 6]		-[(RST - 3) * P - 6]		ns
7	t _{h(XREL-XRDYH)}	Hold time, XRDY high after $\overline{\text{XRE}}$ low	(RST – 3) * P + 2		(RST – 3) * P + 2		ns
9	t _{su(XRDYL-XREL)}	Setup time, XRDY low before $\overline{\text{XRE}}$ low	-[(RST - 3) * P - 6]		-[(RST - 3) * P - 6]		ns
10	t _{h(XREL-XRDYL)}	Hold time, XRDY low after $\overline{\text{XRE}}$ low	(RST – 3) * P + 2		(RST – 3) * P + 2		ns
11	t _{w(XRDYH)}	Pulse width, XRDY high	2P		2P		ns
15	t _{su(XRDYH-XWEL)}	Setup time, XRDY high before XWE low	-[(WST - 3) * P - 6]		-[(WST - 3) * P - 6]		ns
16	t _{h(XWEL-XRDYH)}	Hold time, XRDY high after XWE low	(WST - 3) * P + 2		(WST - 3) * P + 2		ns
18	t _{su(XRDYL-XWEL)}	Setup time, XRDY low before XWE low	-[(WST - 3) * P - 6]		-[(WST - 3) * P - 6]		ns
19	t _{h(XWEL-XRDYL)}	Hold time, XRDY low after $\overline{\text{XWE}}$ low	(WST - 3) * P + 2		(WST - 3) * P + 2		ns

[†] To ensure data setup time, simply program the strobe width wide enough. XRDY is internally synchronized. If XRDY does meet setup or hold time, it may be recognized in the current cycle or the next cycle. Therefore, XRDY can be an asynchronous input.

^{*} RS = Read Setup, RST = Read Strobe, RH = Read Hold, WS = Write Setup, WST = Write Strobe, WH = Write Hold. These parameters are programmed via the XBUS XCE space control registers.

 $[\]$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

The sum of RS and RST (or WS and WST) must be a minimum of 4 in order to use XRDY input to extend strobe width.

SPRS104I - OCTOBER 1999 - REVISED FEBRUARY 2004

EXPANSION BUS ASYNCHRONOUS PERIPHERAL TIMING (CONTINUED)

switching characteristics over recommended operating conditions for asynchronous peripheral cycles †‡§¶ (see Figure 37–Figure 40)

NO.	PARAMETER		C6202-200 C6202-250		C6202B-250 C6202B-300			UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX	
1	t _{osu(SELV-XREL)}	Output setup time, select signals valid to XRE low	RS * P - 2			RS * P - 2			ns
2	t _{oh(XREH-SELIV)}	Output hold time, XRE low to select signals invalid	RH * P – 2			RH * P – 2			ns
5	t _{w(XREL)}	Pulse width, XRE low		RST * P			RST * P		ns
8	t _{d(XRDYH-XREH)}	Delay time, XRDY high to XRE high	3P		4P + 5	3P		4P + 5	ns
12	t _{osu(SELV-XWEL)}	Output setup time, select signals valid to XWE low	WS * P - 2			WS * P - 3			ns
13	toh(XWEH-SELIV)	Output hold time, XWE low to select signals invalid	WH * P – 2			WH * P – 2			ns
14	t _{w(XWEL)}	Pulse width, XWE low		WST*P			WST * P		ns
17	t _d (XRDYH-XWEH)	Delay time, XRDY high to XWE high	3P		4P + 5	3P		4P + 5	ns

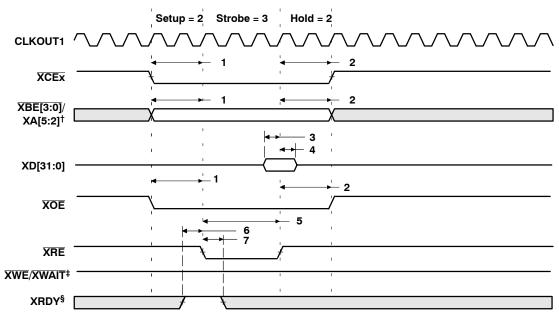
[†] RS = Read Setup, RST = Read Strobe, RH = Read Hold, WS = Write Setup, WST = Write Strobe, WH = Write Hold. These parameters are programmed via the XBUS XCE space control registers.

 $^{^{\}ddagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[§] The sum of RS and RST (or WS and WST) must be a minimum of 4 in order to use XRDY input to extend strobe width.

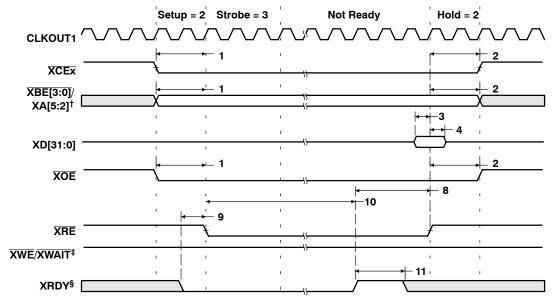
Select signals include: XCEx, XBE[3:0]/XA[5:2], XOE; and for writes, include XD[31:0], with the exception that XCEx can stay active for an additional 7P ns following the end of the cycle.

EXPANSION BUS ASYNCHRONOUS PERIPHERAL TIMING (CONTINUED)



[†] XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during expansion bus asynchronous peripheral accesses.

Figure 37. Expansion Bus Asynchronous Peripheral Read Timing (XRDY Not Used)



[†] XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during expansion bus asynchronous peripheral accesses.

Figure 38. Expansion Bus Asynchronous Peripheral Read Timing (XRDY Used)

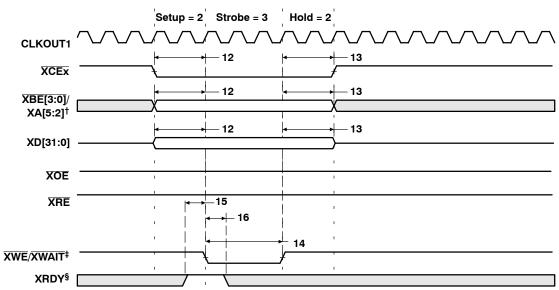
[‡] XWE/XWAIT operates as the write-enable signal XWE during expansion bus asynchronous peripheral accesses.

[§] XRDY operates as active-high ready input during expansion bus asynchronous peripheral accesses.

[‡] XWE/XWAIT operates as the write-enable signal XWE during expansion bus asynchronous peripheral accesses.

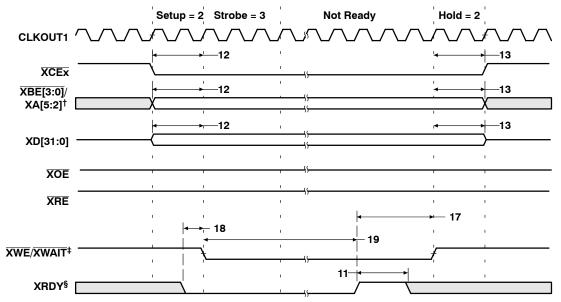
[§] XRDY operates as active-high ready input during expansion bus asynchronous peripheral accesses.

EXPANSION BUS ASYNCHRONOUS PERIPHERAL TIMING (CONTINUED)



- † XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during expansion bus asynchronous peripheral accesses.
- ‡ XWE/XWAIT operates as the write-enable signal XWE during expansion bus asynchronous peripheral accesses.
- § XRDY operates as active-high ready input during expansion bus asynchronous peripheral accesses.

Figure 39. Expansion Bus Asynchronous Peripheral Write Timing (XRDY Not Used)



- † XBE[3:0]/XA[5:2] operate as address signals XA[5:2] during expansion bus asynchronous peripheral accesses.
- ‡ XWE/XWAIT operates as the write-enable signal XWE during expansion bus asynchronous peripheral accesses.
- § XRDY operates as active-high ready input during expansion bus asynchronous peripheral accesses.

Figure 40. Expansion Bus Asynchronous Peripheral Write Timing (XRDY Used)

EXPANSION BUS SYNCHRONOUS HOST-PORT TIMING

timing requirements with external device as bus master (see Figure 41 and Figure 42)

NO.			C6202		C6202		UNIT
			MIN	MAX	MIN	MAX	
1	t _{su(XCSV-XCKIH)}	Setup time, XCS valid before XCLKIN high	3.5		3.5		ns
2	t _{h(XCKIH-XCS)}	Hold time, $\overline{\text{XCS}}$ valid after XCLKIN high	2.8		2.8		ns
3	t _{su(XAS-XCKIH)}	Setup time, XAS valid before XCLKIN high	3.5		3.5		ns
4	t _{h(XCKIH-XAS)}	Hold time, XAS valid after XCLKIN high	2.8		2.8		ns
5	t _{su(XCTL-XCKIH)}	Setup time, XCNTL valid before XCLKIN high	3.5		3.5		ns
6	t _{h(XCKIH-XCTL)}	Hold time, XCNTL valid after XCLKIN high	2.8		2.8		ns
7	t _{su(XWR-XCKIH)}	Setup time, XW/R valid before XCLKIN high†	3.5		3.5		ns
8	t _{h(XCKIH-XWR)}	Hold time, XW/R valid after XCLKIN high †	2.8		2.8		ns
9	t _{su(XBLTV-XCKIH)}	Setup time, XBLAST valid before XCLKIN high [‡]	3.5		3.5		ns
10	t _{h(XCKIH-XBLTV)}	Hold time, XBLAST valid after XCLKIN high [‡]	2.8		2.8		ns
16	t _{su(XBEV-XCKIH)}	Setup time, XBE[3:0]/XA[5:2] valid before XCLKIN high§	3.5		3.5		ns
17	t _{h(XCKIH-XBEV)}	Hold time, XBE[3:0]/XA[5:2] valid after XCLKIN high§	2.8	-	2.8		ns
18	t _{su(XD-XCKIH)}	Setup time, XDx valid before XCLKIN high	3.5		3.5		ns
19	t _{h(XCKIH-XD)}	Hold time, XDx valid after XCLKIN high	2.8	•	2.8	•	ns

[†] XW/R input/output polarity selected at boot.

switching characteristics over recommended operating conditions with external device as bus master[¶] (see Figure 41 and Figure 42)

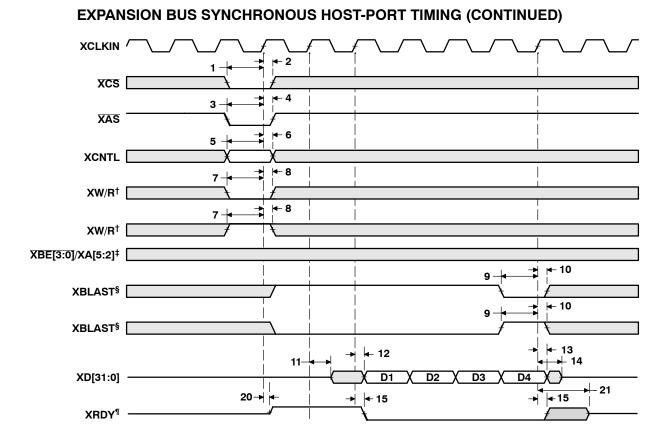
NO.	PARAMETER		C6202-200 C6202-250		C620	UNIT	
			MIN	MAX	MIN	MAX	
11	t _{d(XCKIH-XDLZ)}	Delay time, XCLKIN high to XDx low impedance	0		0		ns
12	t _{d(XCKIH-XDV)}	Delay time, XCLKIN high to XDx valid		16.5		4P – 0.5	ns
13	t _{d(XCKIH-XDIV)}	Delay time, XCLKIN high to XDx invalid	5		3		ns
14	t _{d(XCKIH-XDHZ)}	Delay time, XCLKIN high to XDx high impedance		4P		4P	ns
15	t _{d(XCKIH-XRY)}	Delay time, XCLKIN high to XRDY invalid#	5	16.5	3	4P – 0.5	ns
20	t _{d(XCKIH-XRYLZ)}	Delay time, XCLKIN high to XRDY low impedance	5	16.5	3	4P – 0.5	ns
21	t _{d(XCKIH-XRYHZ)}	Delay time, XCLKIN high to XRDY high impedance#	2P + 5	3P + 16.5	2P + 3	7P + 0.5	ns

 $[\]P$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[‡] XBLAST input polarity selected at boot

[§] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

[#] XRDY operates as active-low ready input/output during host-port accesses.



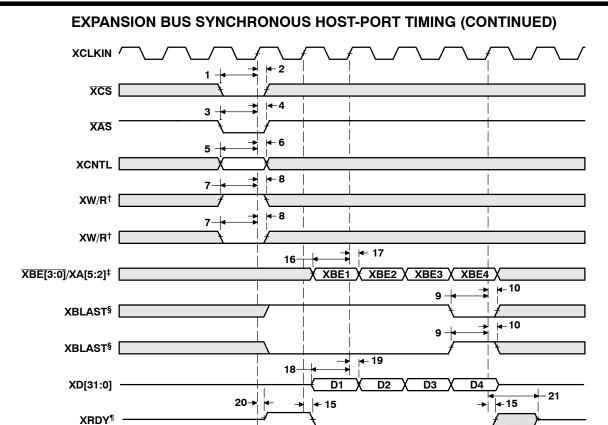
[†] XW/R input/output polarity selected at boot

Figure 41. External Host as Bus Master—Read

[‡] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

[§] XBLAST input polarity selected at boot

[¶] XRDY operates as active-low ready input/output during host-port accesses.



[†] XW/R input/output polarity selected at boot

Figure 42. External Host as Bus Master—Write

[‡] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

[§] XBLAST input polarity selected at boot

[¶] XRDY operates as active-low ready input/output during host-port accesses.

EXPANSION BUS SYNCHRONOUS HOST-PORT TIMING (CONTINUED)

timing requirements with C62x[™] as bus master (see Figure 43, Figure 44, and Figure 45)

NO.			C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT
			MIN	MAX	MIN	MAX	
9	t _{su(XDV-XCKIH)}	Setup time, XDx valid before XCLKIN high	3.5		3.5		ns
10	t _{h(XCKIH-XDV)}	Hold time, XDx valid after XCLKIN high	2.8		2.8		ns
11	t _{su(XRY-XCKIH)}	Setup time, XRDY valid before XCLKIN high [†]	3.5		3.5		ns
12	t _{h(XCKIH-XRY)}	Hold time, XRDY valid after XCLKIN high [†]	2.8		2.8		ns
14	t _{su(XBFF-XCKIH)}	Setup time, XBOFF valid before XCLKIN high	3.5		3.5		ns
15	t _{h(XCKIH-XBFF)}	Hold time, XBOFF valid after XCLKIN high	2.8		2.8		ns

[†] XRDY operates as active-low ready input/output during host-port accesses.

switching characteristics over recommended operating conditions with C62x as bus master (see Figure 43, Figure 44, and Figure 45)

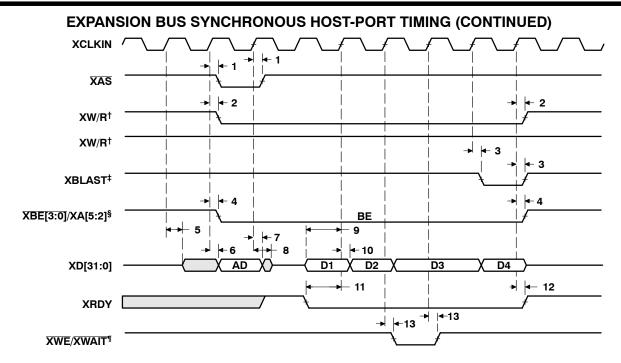
NO.	PARAMETER			C6202-200 C6202-250		C6202B-250 C6202B-300		
			MIN	MAX	MIN	MAX		
1	t _{d(XCKIH-XASV)}	Delay time, XCLKIN high to XAS valid	5	16.5	3	4P – 0.5	ns	
2	t _{d(XCKIH-XWRV)}	Delay time, XCLKIN high to XW/R valid [‡]	5	16.5	3	4P – 0.5	ns	
3	t _{d(XCKIH-XBLTV)}	Delay time, XCLKIN high to XBLAST valid§	5	16.5	3	4P – 0.5	ns	
4	t _{d(XCKIH-XBEV)}	Delay time, XCLKIN high to XBE[3:0]/XA[5:2] valid [¶]	5	16.5	3	4P – 0.5	ns	
5	t _{d(XCKIH-XDLZ)}	Delay time, XCLKIN high to XDx low impedance	0		0		ns	
6	t _{d(XCKIH-XDV)}	Delay time, XCLKIN high to XDx valid		16.5		4P – 0.5	ns	
7	t _{d(XCKIH-XDIV)}	Delay time, XCLKIN high to XDx invalid	5		3		ns	
8	t _{d(XCKIH-XDHZ)}	Delay time, XCLKIN high to XDx high impedance		4P		4P	ns	
13	t _{d(XCKIH-XWTV)}	Delay time, XCLKIN high to XWE/XWAIT valid#	5	16.5	3	4P – 0.5	ns	

[‡] XW/R input/output polarity selected at boot.

[§] XBLAST output polarity is always active low.

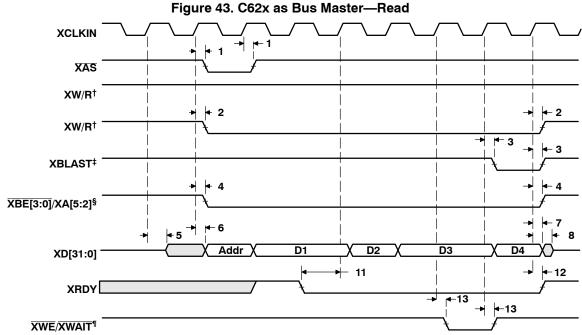
TBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

[#] XWE/XWAIT operates as XWAIT output signal during host-port accesses.



[†] XW/R input/output polarity selected at boot

TWE/XWAIT operates as XWAIT output signal during host-port accesses.



[†] XW/R input/output polarity selected at boot

Figure 44. C62x as Bus Master—Write



[‡] XBLAST output polarity is always active low.

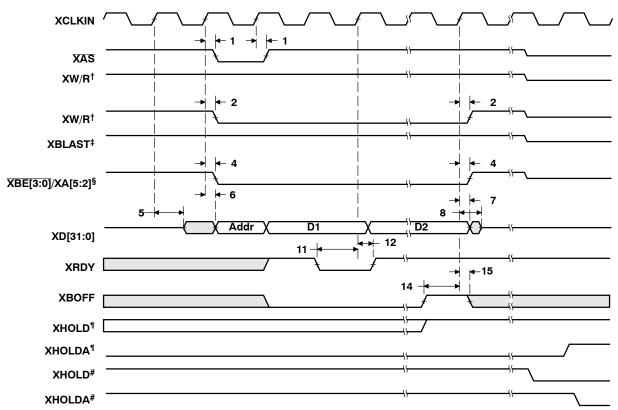
[§] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

[‡] XBLAST output polarity is always active low.

[§] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

TWE/XWAIT operates as XWAIT output signal during host-port accesses.

EXPANSION BUS SYNCHRONOUS HOST-PORT TIMING (CONTINUED)



[†] XW/R input/output polarity selected at boot

Figure 45. C62x as Bus Master—BOFF Operation

[‡] XBLAST output polarity is always active low.

[§] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

[¶] Internal arbiter enabled

[#] Internal arbiter disabled

 $[\]parallel$ This diagram illustrates XBOFF timing. Bus arbitration timing is shown in Figure 48 and Figure 49.

EXPANSION BUS ASYNCHRONOUS HOST-PORT TIMING

timing requirements with external device as asynchronous bus master † (see Figure 46 and Figure 47)

NO.			C6202 C6202		C6202B-250 C6202B-300		UNIT
			MIN	MAX	MIN	MAX	
1	t _{w(XCSL)}	Pulse duration, XCS low	4P		4P		ns
2	t _{w(XCSH)}	Pulse duration, XCS high	4P		4P		ns
3	t _{su(XSEL-XCSL)}	Setup time, expansion bus select signals ‡ valid before $\overline{\text{XCS}}$ low	1		1		ns
4	t _{h(XCSL-XSEL)}	Hold time, expansion bus select signals ‡ valid after $\overline{\text{XCS}}$ low	3		3.4		ns
10	t _{h(XRYL-XCSL)}	Hold time, $\overline{\text{XCS}}$ low after XRDY low	P + 1.5		P + 1.5		ns
11	t _{su(XBEV-XCSH)}	Setup time, $\overline{\text{XBE}[3:0]}/\text{XA}[5:2]$ valid before $\overline{\text{XCS}}$ high§	1		1		ns
12	t _{h(XCSH-XBEV)}	Hold time, XBE[3:0]/XA[5:2] valid after XCS high§	3		3		ns
13	t _{su(XDV-XCSH)}	Setup time, XDx valid before $\overline{\text{XCS}}$ high	1		1		ns
14	t _{h(XCSH-XDV)}	Hold time, XDx valid after XCS high	3		3		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions with external device as asynchronous bus master[†] (see Figure 46 and Figure 47)

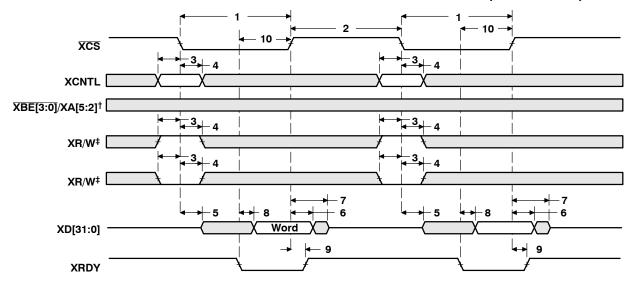
NO.	PARAMETER		C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT
		MIN N		MAX	MIN	MAX	
5	t _{d(XCSL-XDLZ)}	Delay time, XCS low to XDx low impedance	0		0		ns
6	t _{d(XCSH-XDIV)}	Delay time, XCS high to XDx invalid	0	12	0	12	ns
7	t _{d(XCSH-XDHZ)}	Delay time, XCS high to XDx high impedance		4P		4P	ns
8	t _{d(XRYL-XDV)}	Delay time, XRDY low to XDx valid	-4	1	-4	1.8	ns
9	t _{d(XCSH-XRYH)}	Delay time, XCS high to XRDY high	0	12	-1	12	ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[‡] Expansion bus select signals include XCNTL and XR/W.

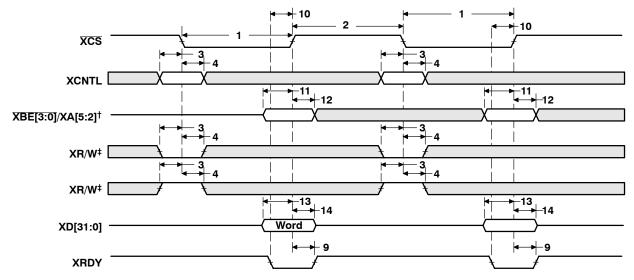
[§] XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

EXPANSION BUS ASYNCHRONOUS HOST-PORT TIMING (CONTINUED)



 $^{^{\}dagger}$ $\overline{\text{XBE}[3:0]}$ /XA[5:2] operate as byte-enables $\overline{\text{XBE}[3:0]}$ during host-port accesses.

Figure 46. External Device as Asynchronous Master—Read



 $^{^\}dagger$ XBE[3:0]/XA[5:2] operate as byte-enables XBE[3:0] during host-port accesses.

Figure 47. External Device as Asynchronous Master—Write

[‡] XW/R input/output polarity selected at boot

[‡] XW/R input/output polarity selected at boot

XHOLD/XHOLDA TIMING

timing requirements for expansion bus arbitration (internal arbiter enabled)[†] (see Figure 48)

NO.			C6202 C6202 C6202 C6202	-250 3-250	UNIT
			MIN	MAX	
3	t _{oh(XHDAH-XHDH)}	Output hold time, XHOLD high after XHOLDA high	Р		ns

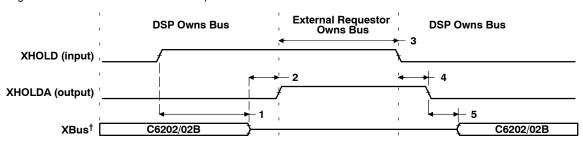
[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for expansion bus arbitration (internal arbiter enabled)^{†‡} (see Figure 48)

NO.		PARAMETER	C6202 C6202 C6202 C6202	2-250 B-250	UNIT
			MIN	MAX	
1	t _{d(XHDH-XBHZ)}	Delay time, XHOLD high to XBus high impedance	3P	§	ns
2	t _{d(XBHZ-XHDAH)}	Delay time, XBus high impedance to XHOLDA high	0	2P	ns
4	t _{d(XHDL-XHDAL)}	Delay time, XHOLD low to XHOLDA low	3P		ns
5	t _{d(XHDAL-XBLZ)}	Delay time, XHOLDA low to XBus low impedance	0	2P	ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[§] All pending XBus transactions are allowed to complete before XHOLDA is asserted.



[†] XBus consists of XBE[3:0]/XA[5:2], XAS, XW/R, and XBLAST.

Figure 48. Expansion Bus Arbitration—Internal Arbiter Enabled

[‡] XBus consists of XBE[3:0]/XA[5:2], XAS, XW/R, and XBLAST.

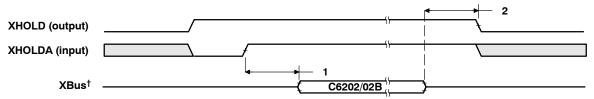
XHOLD/XHOLDA TIMING (CONTINUED)

switching characteristics over recommended operating conditions for expansion bus arbitration (internal arbiter disabled)[†] (see Figure 49)

NO.		PARAMETER	C620	02-200 02-250 2B-250 2B-300	UNIT
	M		MIN	MAX	
1	t _{d(XHDAH-XBLZ)}	Delay time, XHOLDA high to XBus low impedance [‡]	2P	2P + 10	ns
2	t _{d(XBHZ-XHDL)}	Delay time, XBus high impedance to XHOLD low [‡]	0	2P	ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

[‡] XBus consists of XBE[3:0]/XA[5:2], XAS, XW/R, and XBLAST.



 $^{^{\}dagger}$ XBus consists of $\overline{\text{XBE}[3:0]}/\text{XA}[5:2],$ $\overline{\text{XAS}},$ XW/R, and XBLAST.

Figure 49. Expansion Bus Arbitration—Internal Arbiter Disabled

MULTICHANNEL BUFFERED SERIAL PORT TIMING

timing requirements for McBSP^{†‡} (see Figure 50)

NO.			C6202-200 C6202-250		C6202B-250 C6202B-300		UNIT	
							MAX	
2	t _{c(CKRX)}	Cycle time, CLKR/X	CLKR/X ext	2P§		2P§		ns
3	t _{w(CKRX)}	Pulse duration, CLKR/X high or CLKR/X low	CLKR/X ext	P-1 [¶]		P-1 [¶]		ns
-		Out a line as to real FOR bish hafe as OLKR to	CLKR int	9		9		
5	t _{su(FRH-CKRL)}	Setup time, external FSR high before CLKR low	CLKR ext	2		2	ns	ns
		11.11.	CLKR int	6		6		
6	^t h(CKRL-FRH)	Hold time, external FSR high after CLKR low	CLKR ext	3		3		ns
_			CLKR int	8		8		
7	t _{su(DRV-CKRL)}	Setup time, DR valid before CLKR low	CLKR ext	0.5		0.5		ns
		THE PROPERTY OF STREET	CLKR int	3		3		
8	th(CKRL-DRV)	Hold time, DR valid after CLKR low	CLKR ext	4		4.5		ns
40		Oct of the control FOV bab bafa a OHOV b	CLKX int	9		9		
10	t _{su(FXH-CKXL)} Set	Setup time, external FSX high before CLKX low	CLKX ext	2		2		ns
	1.	Hald to a second FOV halo offer OHOV ha	CLKX int	6		6		
11	t _{h(CKXL-FXH)}	Hold time, external FSX high after CLKX low	CLKX ext	3		4		ns

[†] CLKRP = CLKXP = FSRP = FSXP = 0. If the polarity of any of the signals is inverted, then the timing references of that signal are also inverted. ‡ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

The maximum bit rate for the C6202/02B device is 100 Mbps or CPU/2 (the slower of the two). Care must be taken to ensure that the AC timings specified in this data sheet are met. The maximum bit rate for McBSP-to-McBSP communications is 100 MHz; therefore, the minimum CLKR/X clock cycle is either twice the CPU cycle time (2P), or 10 ns (100 MHz), whichever value is larger. For example, when running parts at 250 MHz (P = 4 ns), use 10 ns as the minimum CLKR/X clock cycle (by setting the appropriate CLKGDV ratio or external clock source). When running parts at 100 MHz (P = 10 ns), use 2P = 20 ns (50 MHz) as the minimum CLKR/X clock cycle. The maximum bit rate for McBSP-to-McBSP communications applies when the serial port is a master of the clock and frame syncs (with CLKR connected to CLKX, FSR connected to FSX, CLKXM = FSXM = 1, and CLKRM = FSRM = 0) in data delay 1 or 2 mode (R/XDATDLY = 01b or 10b) and the other device the McBSP communicates to is a slave.

The minimum CLKR/X pulse duration is either (P-1) or 4 ns, whichever is larger. For example, when running parts at 250 MHz (P=4 ns), use 4 ns as the minimum CLKR/X pulse duration. When running parts at 100 MHz (P=10 ns), use (P-1)=9 ns as the minimum CLKR/X pulse duration.

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

switching characteristics over recommended operating conditions for McBSP^{†‡} (see Figure 50)

NO.		PARAMETER			C6202-200 C6202-250		C6202B-250 C6202B-300	
				MIN	MAX	MIN	MAX	
1	t _{d(CKSH-CKRXH)}	Delay time, CLKS high to CLKR/X high for internal CLKR/X generated from CLKS input		4	16	4	16	ns
2	t _{c(CKRX)}	Cycle time, CLKR/X	CLKR/X int	2P ^{§¶}		2P ^{§¶}		ns
3	t _{w(CKRX)}	Pulse duration, CLKR/X high or CLKR/X low	CLKR/X int	C – 1#	C + 1#	C – 1#	C + 1#	ns
4	t _{d(CKRH-FRV)}	Delay time, CLKR high to internal FSR valid	CLKR int	-2	3	-2	3	ns
		Dolov time CLVV high to internal ESV valid	CLKX int	-2	3	-2	3	
9	t _d (CKXH-FXV)	Delay time, CLKX high to internal FSX valid	CLKX ext	3	9	2	9	ns
40		Disable time, DX high impedance following last	CLKX int	-1	5	-1	5	
12	^t dis(CKXH-DXHZ)	data bit from CLKX high	CLKX ext	2	9	2	9	ns
40		Data Care OHOVERS DV and	CLKX int	-1	4	-0.5	3	
13	t _d (CKXH-DXV)	Delay time, CLKX high to DX valid	CLKX ext	2	11	2	11	ns
44		Dolay linic, I OX high to DX valid ONE applies	FSX int	-1	5	-1	5	
14	t _d (FXH-DXV)	when in data delay 0 (XDATDLY = 00b) mode.	FSX ext	0	10	0	10	ns

[†] CLKRP = CLKXP = FSRP = FSXP = 0. If the polarity of any of the signals is inverted, then the timing references of that signal are also inverted.

 $^{\#}C = HorL$

S = sample rate generator input clock = P if CLKSM = 1 (P = 1/CPU clock frequency)

= sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)

H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

CLKGDV should be set appropriately to ensure the McBSP bit rate does not exceed the 100-MHz limit.

[‡] Minimum delay times also represent minimum output hold times.

[§] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

The maximum bit rate for the C6202/02B device is 100 Mbps or CPU/2 (the slower of the two). Care must be taken to ensure that the AC timings specified in this data sheet are met. The maximum bit rate for McBSP-to-McBSP communications is 100 MHz; therefore, the minimum CLKR/X clock cycle is either twice the CPU cycle time (2P), or 10 ns (100 MHz), whichever value is larger. For example, when running parts at 250 MHz (P = 4 ns), use 10 ns as the minimum CLKR/X clock cycle (by setting the appropriate CLKGDV ratio or external clock source). When running parts at 100 MHz (P = 10 ns), use 2P = 20 ns (50 MHz) as the minimum CLKR/X clock cycle. The maximum bit rate for McBSP-to-McBSP communications applies when the serial port is a master of the clock and frame syncs (with CLKR connected to CLKX, FSR connected to FSX, CLKXM = FSXM = 1, and CLKRM = FSRM = 0) in data delay 1 or 2 mode (R/XDATDLY = 01b or 10b) and the other device the McBSP communicates to is a slave.

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

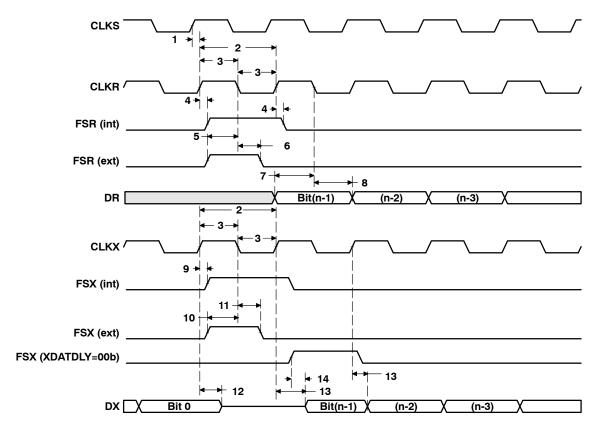


Figure 50. McBSP Timings

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

timing requirements for FSR when GSYNC = 1 (see Figure 51)

NO.			C6202 C6202 C6202 MIN	2-250 B-250	UNIT
1	t _{su(FRH-CKSH)}	Setup time, FSR high before CLKS high	4		ns
2	t _{h(CKSH-FRH)}	Hold time, FSR high after CLKS high	4		ns

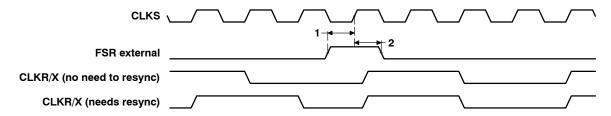


Figure 51. FSR Timing When GSYNC = 1

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

timing requirements for McBSP as SPI master or slave: CLKSTP = 10b, CLKXP = 0^{†‡} (see Figure 52)

NO.				C620	02-200 02-250 2B-250 2B-300	UNIT	
			MAS	ΓER	SLAVE		!
			MIN	MAX	MIN	MAX	
4	t _{su(DRV-CKXL)}	Setup time, DR valid before CLKX low	12		2 – 3P		ns
5	t _{h(CKXL-DRV)}	Hold time, DR valid after CLKX low	4		5 + 6P		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for McBSP as SPI master or slave: CLKSTP = 10b, CLKXP = $0^{\dagger \ddagger}$ (see Figure 52)

NO.	PARAMETER			C6202-200 C6202-250 C6202B-250 C6202B-300					
			MAS	TER§	SL	AVE			
			MIN	MAX	MIN	MAX			
1	t _{h(CKXL-FXL)}	Hold time, FSX low after CLKX low [¶]	T – 2	T + 3			ns		
2	t _{d(FXL-CKXH)}	Delay time, FSX low to CLKX high#	L – 2	L + 3			ns		
3	t _{d(CKXH-DXV)}	Delay time, CLKX high to DX valid	-3	4	3P + 4	5P + 17	ns		
6	t _{dis(CKXL-DXHZ)}	Disable time, DX high impedance following last data bit from CLKX low	L – 2	L + 3			ns		
7	t _{dis(FXH-DXHZ)}	Disable time, DX high impedance following last data bit from FSX high			P+3	3P + 17	ns		
8	t _{d(FXL-DXV)}	Delay time, FSX low to DX valid			2P + 2	4P + 17	ns		

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

T = CLKX period = (1 + CLKGDV) * S

H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

CLKGDV should be set appropriately to ensure the McBSP bit rate does not exceed the 100-MHz limit.

CLKXM = FSXM = 1, CLKRM = FSRM = 0 for master McBSP

CLKXM = CLKRM = FSXM = FSRM = 0 for slave McBSP



[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[§] S = sample rate generator input clock = P if CLKSM = 1 (P = 1/CPU clock frequency)

⁼ sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)

FSRP = FSXP = 1. As a SPI master, FSX is inverted to provide active-low slave-enable output. As a slave, the active-low signal input on FSX and FSR is inverted before being used internally.

[#] FSX should be low before the rising edge of clock to enable slave devices and then begin a SPI transfer at the rising edge of the master clock (CLKX).

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

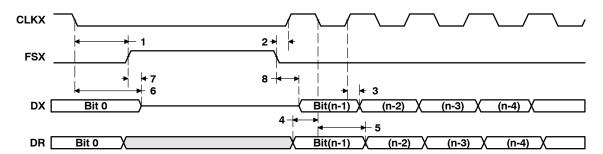


Figure 52. McBSP Timing as SPI Master or Slave: CLKSTP = 10b, CLKXP = 0

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

timing requirements for McBSP as SPI master or slave: CLKSTP = 11b, CLKXP = 0^{†‡} (see Figure 53)

NO.				C6202-200 C6202-250 C6202B-250 C6202B-300			UNIT
			MAS	ΓER	SLAVE		
			MIN	MAX	MIN	MAX	
4	t _{su(DRV-CKXH)}	Setup time, DR valid before CLKX high	12		2 – 3P		ns
5	t _{h(CKXH-DRV)}	Hold time, DR valid after CLKX high	4		5 + 6P		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for McBSP as SPI master or slave: CLKSTP = 11b, CLKXP = $0^{\dagger \ddagger}$ (see Figure 53)

NO.	PARAMETER			C6202-200 C6202-250 C6202B-250 C6202B-300					
			MAS	TER§	SL	AVE			
			MIN	MAX	MIN	MAX			
1	t _{h(CKXL-FXL)}	Hold time, FSX low after CLKX low [¶]	L – 2	L + 3			ns		
2	t _{d(FXL-CKXH)}	Delay time, FSX low to CLKX high#	T – 2	T + 3			ns		
3	t _{d(CKXL-DXV)}	Delay time, CLKX low to DX valid	-2	4	3P + 4	5P + 17	ns		
6	t _{dis(CKXL-DXHZ)}	Disable time, DX high impedance following last data bit from CLKX low	-2	4	3P + 3	5P + 17	ns		
7	t _{d(FXL-DXV)}	Delay time, FSX low to DX valid	H – 2	H + 4	2P + 2	4P + 17	ns		

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

The maximum transfer rate for SPI mode is limited to the above AC timing constraints.

CLKXM = FSXM = 1, CLKRM = FSRM = 0 for master McBSP

CLKXM = CLKRM = FSXM = FSRM = 0 for slave McBSP



[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[§] S = sample rate generator input clock = P if CLKSM = 1 (P = 1/CPU clock frequency)

⁼ sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)

T = CLKX period = (1 + CLKGDV) * S

FSRP = FSXP = 1. As a SPI master, FSX is inverted to provide active-low slave-enable output. As a slave, the active-low signal input on FSX and FSR is inverted before being used internally.

[#] FSX should be low before the rising edge of clock to enable slave devices and then begin a SPI transfer at the rising edge of the master clock (CLKX).

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

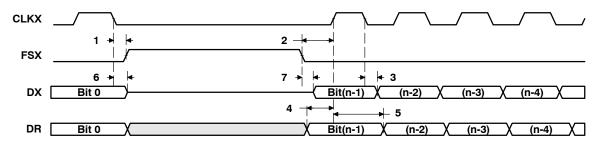


Figure 53. McBSP Timing as SPI Master or Slave: CLKSTP = 11b, CLKXP = 0

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

timing requirements for McBSP as SPI master or slave: CLKSTP = 10b, CLKXP = 1^{†‡} (see Figure 54)

NO.				C620	02-200 02-250 2B-250 2B-300		UNIT
			MAS	ΓER	ER SLAVE		
			MIN	MAX	MIN	MAX	
4	t _{su(DRV-CKXH)}	Setup time, DR valid before CLKX high	12		2 – 3P		ns
5	t _{h(CKXH-DRV)}	Hold time, DR valid after CLKX high	4		5 + 6P		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for McBSP as SPI master or slave: CLKSTP = 10b, CLKXP = $1^{\dagger \ddagger}$ (see Figure 54)

	PARAMETER		C6202-200 C6202-250				C6202B-250 C6202B-300				UNIT
NO.			MASTER§		SL	SLAVE		MASTER§		SLAVE	
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
1	t _{h(CKXH-FXL)}	Hold time, FSX low after CLKX high [¶]	T – 2	T + 3			T – 2	T + 3			ns
2	t _{d(FXL-CKXL)}	Delay time, FSX low to CLKX low#	H – 2	H + 3			H – 2	H + 3			ns
3	t _{d(CKXL-DXV)}	Delay time, CLKX low to DX valid	-2	4	3P + 4	5P + 17	-3	4	3P + 4	5P + 17	ns
6	t _{dis(CKXH-DXHZ)}	Disable time, DX high impedance following last data bit from CLKX high	H – 2	H + 3			H – 2	H + 3			ns
7	t _{dis(FXH-DXHZ)}	Disable time, DX high impedance following last data bit from FSX high			P + 3	3P + 17			P + 3	3P + 17	ns
8	t _{d(FXL-DXV)}	Delay time, FSX low to DX valid			2P + 2	4P + 17			2P + 2	4P + 17	ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

CLKXM = FSXM = 1, CLKRM = FSRM = 0 for master McBSP

CLKXM = CLKRM = FSXM = FSRM = 0 for slave McBSP



[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[§] S = sample rate generator input clock = P if CLKSM = 1 (P = 1/CPU clock frequency)

⁼ sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)

T = CLKX period = (1 + CLKGDV) * S

H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even

L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even

The maximum transfer rate for SPI mode is limited to the above AC timing constraints.

FSRP = FSXP = 1. As a SPI master, FSX is inverted to provide active-low slave-enable output. As a slave, the active-low signal input on FSX and FSR is inverted before being used internally.

[#] FSX should be low before the rising edge of clock to enable slave devices and then begin a SPI transfer at the rising edge of the master clock (CLKX).

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

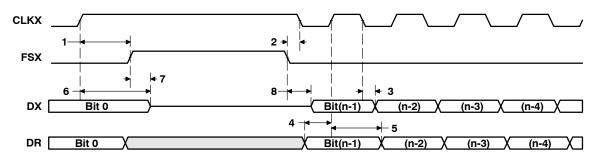


Figure 54. McBSP Timing as SPI Master or Slave: CLKSTP = 10b, CLKXP = 1

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

timing requirements for McBSP as SPI master or slave: CLKSTP = 11b, CLKXP = 1^{†‡} (see Figure 55)

NO.				C620	02-200 02-250 2B-250 2B-300	UNIT	
			MAS	ΓER	SLAVE		!
			MIN	MAX	MIN	MAX	
4	t _{su(DRV-CKXL)}	Setup time, DR valid before CLKX low	12		2 – 3P		ns
5	t _{h(CKXL-DRV)}	Hold time, DR valid after CLKX low	4		5 + 6P		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for McBSP as SPI master or slave: CLKSTP = 11b, CLKXP = $1^{\dagger \ddagger}$ (see Figure 55)

NO.			UNIT				
			MAS	TER§	SL	AVE	
			MIN	MAX	MIN	MAX	
1	t _{h(CKXH-FXL)}	Hold time, FSX low after CLKX high [¶]	H – 2	H + 3			ns
2	t _{d(FXL-CKXL)}	Delay time, FSX low to CLKX low#	T – 2	T + 2			ns
3	t _{d(CKXH-DXV)}	Delay time, CLKX high to DX valid	-3	4	3P + 4	5P + 17	ns
6	t _{dis(CKXH-DXHZ)}	Disable time, DX high impedance following last data bit from CLKX high	-2	4	3P + 3	5P + 17	ns
7	t _{d(FXL-DXV)}	Delay time, FSX low to DX valid	L – 2	L + 5	2P + 2	4P + 17	ns

 $^{^\}dagger$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

H = CLKX high pulse width = (CLKGDV/2 + 1) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

L = CLKX low pulse width = (CLKGDV/2) * S if CLKGDV is even

= (CLKGDV + 1)/2 * S if CLKGDV is odd or zero

CLKGDV should be set appropriately to ensure the McBSP bit rate does not exceed the 100-MHz limit.

CLKXM = FSXM = 1, CLKRM = FSRM = 0 for master McBSP

CLKXM = CLKRM = FSXM = FSRM = 0 for slave McBSP



[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[‡] For all SPI slave modes, CLKG is programmed as 1/2 of the CPU clock by setting CLKSM = CLKGDV = 1.

[§] S = sample rate generator input clock = P if CLKSM = 1 (P = 1/CPU clock frequency)

⁼ sample rate generator input clock = P_clks if CLKSM = 0 (P_clks = CLKS period)

T = CLKX period = (1 + CLKGDV) * S

FSRP = FSXP = 1. As a SPI master, FSX is inverted to provide active-low slave-enable output. As a slave, the active-low signal input on FSX and FSR is inverted before being used internally.

[#] FSX should be low before the rising edge of clock to enable slave devices and then begin a SPI transfer at the rising edge of the master clock (CLKX).

MULTICHANNEL BUFFERED SERIAL PORT TIMING (CONTINUED)

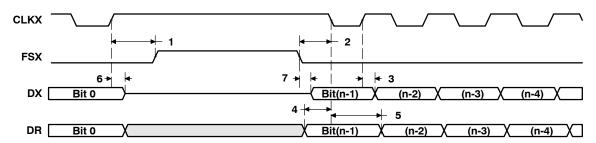


Figure 55. McBSP Timing as SPI Master or Slave: CLKSTP = 11b, CLKXP = 1

DMAC, TIMER, POWER-DOWN TIMING

switching characteristics over recommended operating conditions for DMAC outputs[†] (see Figure 56)

NO.	PARAMETER	C6202 C6202 C6202I C6202I	2-250 B-250	UNIT
		MIN	MAX	
1	t _{w(DMACH)} Pulse duration, DMAC high	2P-3		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.



Figure 56. DMAC Timing

timing requirements for timer inputs[†] (see Figure 57)

NO.		C6202 C6202 C6202 MIN	2-250 B-250	UNIT
1	t _{w(TINPH)} Pulse duration, TINP high	2P		ns
2	t _{w(TINPL)} Pulse duration, TINP low	2P		ns

[†] P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

switching characteristics over recommended operating conditions for timer outputs[†] (see Figure 57)

NO.		PARAMETER	C6202 C6202 C6202 C6202 MIN	2-250 B-250	UNIT
3	t _{w(TOUTH)}	Pulse duration, TOUT high	2P-3		ns
4	t _{w(TOUTL)}	Pulse duration, TOUT low	2P-3		ns

 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

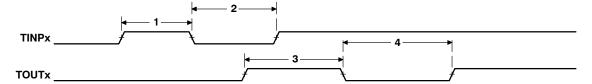
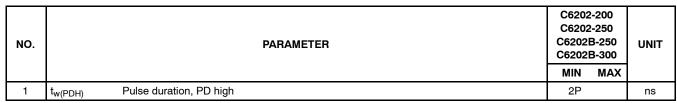


Figure 57. Timer Timing

DMAC, TIMER, POWER-DOWN TIMING (CONTINUED)

switching characteristics over recommended operating conditions for power-down outputs[†] (see Figure 58)



 $^{^{\}dagger}$ P = 1/CPU clock frequency in ns. For example, when running parts at 250 MHz, use P = 4 ns.

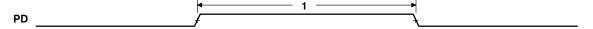


Figure 58. Power-Down Timing

JTAG TEST-PORT TIMING

timing requirements for JTAG test port (see Figure 59)

NO.			C6202 C6202 C6202E C6202E	-250 3-250	UNIT
1	t _{c(TCK)}	Cycle time, TCK	35		ns
3	t _{su(TDIV-TCKH)}	Setup time, TDI/TMS/TRST valid before TCK high	11		ns
4	t _{h(TCKH-TDIV)}	Hold time, TDI/TMS/TRST valid after TCK high	9		ns

switching characteristics over recommended operating conditions for JTAG test port (see Figure 59)

NO.	PARAMETER	C6202 C6202		C6202B-250 C6202B-300		UNIT
		MIN	MAX	MIN	MAX	
2	t _{d(TCKL-TDOV)} Delay time, TCK low to TDO valid	-4.5	12	-4.5	13.5	ns

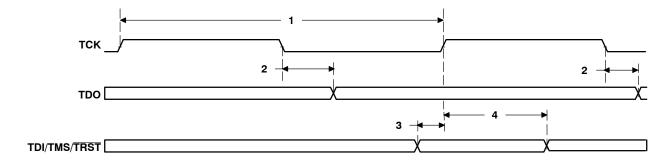
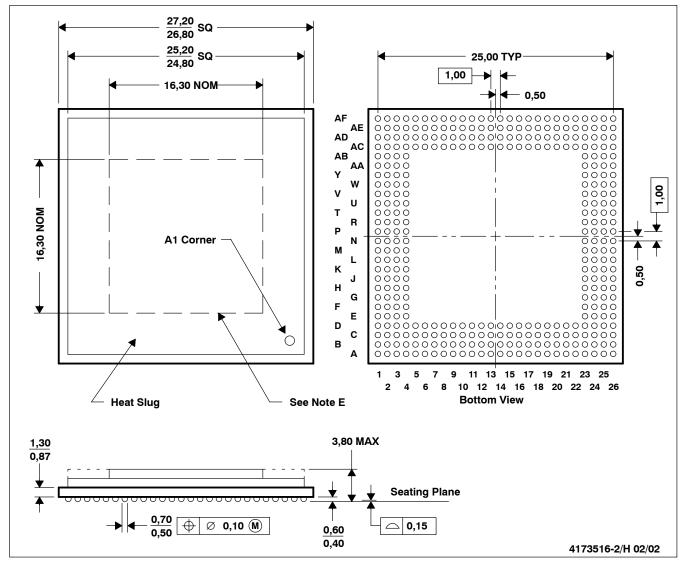


Figure 59. JTAG Test-Port Timing

MECHANICAL DATA

GJL (S-PBGA-N352) [C6202 only]

PLASTIC BALL GRID ARRAY



- NOTES: A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Thermally enhanced plastic package with heat slug (HSL)
 - D. Flip chip application only
 - E. Possible protrusion in this area, but within 3,50 max package height specification
 - Falls within JEDEC MO-151/AAL-1

thermal resistance characteristics (S-PBGA package) [C6202 only]

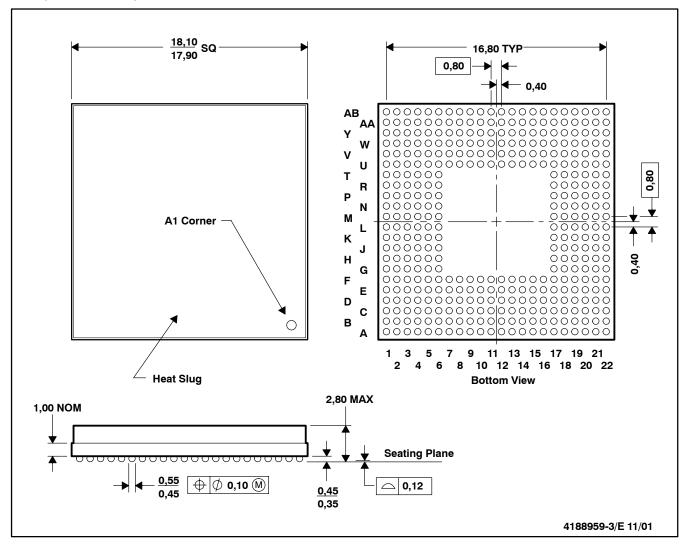
NO		°C/W	Air Flow m/s [†]
1	$R\Theta_{JC}$ Junction-to-case	0.47	N/A
2	RΘ _{JA} Junction-to-free air	14.2	0.00
3	R⊖ _{JA} Junction-to-free air	12.3	0.50
4	R⊖ _{JA} Junction-to-free air	10.9	1.00
5	RΘ _{JA} Junction-to-free air	9.3	2.00

[†] m/s = meters per second

MECHANICAL DATA

GLS (S-PBGA-N384)

PLASTIC BALL GRID ARRAY



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- Thermally enhanced plastic package with heat slug (HSL)
- Flip chip application only

thermal resistance characteristics (S-PBGA package)) [C6202 only]

NO		°C/W	Air Flow m/s†
1	$R\Theta_{JC}$ Junction-to-case	0.85	N/A
2	$R\Theta_{JA}$ Junction-to-free air	21.6	0.00
3	$R\Theta_{JA}$ Junction-to-free air	18.0	0.50
4	$R\Theta_{JA}$ Junction-to-free air	15.5	1.00
5	$R\Theta_{JA}$ Junction-to-free air	12.8	2.00

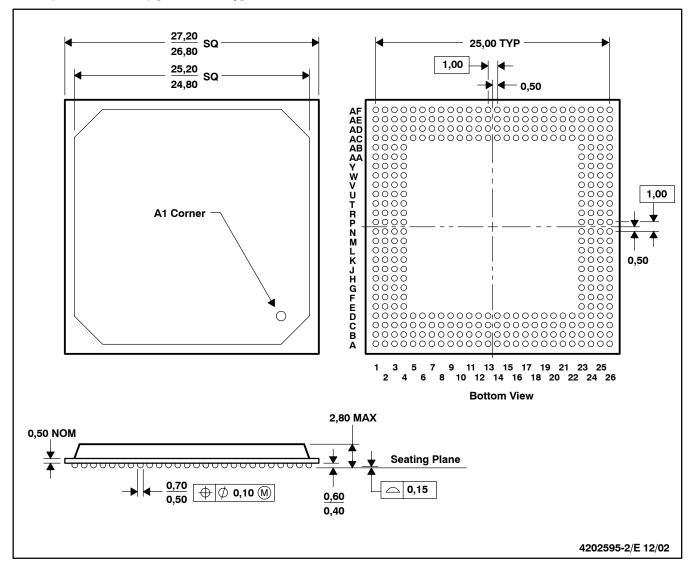
[†] m/s = meters per second



MECHANICAL DATA

GNZ (S-PBGA-N352) [C6202B only]

PLASTIC BALL GRID ARRAY



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Flip chip application only.
- D. Substrate color may vary.

thermal resistance characteristics (S-PBGA package) [C6202B only]

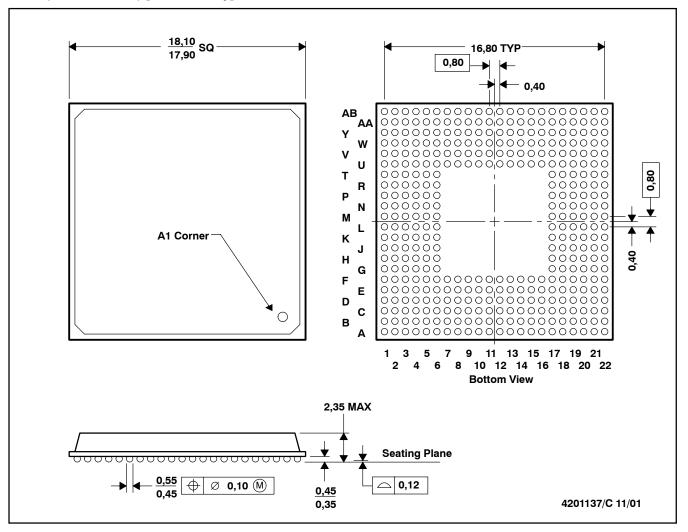
NO			°C/W	Air Flow m/s†
1	RΘ _{JC} Junction	on-to-case	6.35	N/A
2	RΘ _{JA} Junction	on-to-free air	20.0	0.00
3	RΘ _{JA} Junction	on-to-free air	17.0	0.50
4	RΘ _{JA} Junction	on-to-free air	16.3	1.00
5	RΘ _{JA} Junction	on-to-free air	15.2	2.00

[†] m/s = meters per second

MECHANICAL DATA

GNY (S-PBGA-N384) [C6202B only]

PLASTIC BALL GRID ARRAY



NOTES: A. All linear dimensions are in millimeters.

- This drawing is subject to change without notice.
- Flip chip application only
- Substrate color may vary

thermal resistance characteristics (S-PBGA package) [C6202B only]

NO		(°C/W)	Air Flow m/s†
1	$R\Theta_{JC}$ Junction-to-case	6.27	N/A
2	$R\Theta_{JA}$ Junction-to-free air	17.6	0.0
3	$R\Theta_{JA}$ Junction-to-free air	13.9	0.5
4	$R\Theta_{JA}$ Junction-to-free air	13.1	1.0
5	$R\Theta_{JA}$ Junction-to-free air	11.9	2.0

[†] m/s = meters per second



REVISION HISTORY

This data sheet revision history highlights the technical changes made to the SPRS104H device-specific data sheet to make it an SPRS104I revision.

SCOPE: This document has been revised to add information pertaining to power-down mode logic.

PAGE(S) NO.	ADDITIONS/CHANGES/DELETIONS
12	memory map summary: Changed the document reference in the last sentence of the paragraph.
13	peripheral register descriptions: Updated the information regarding the document reference.
18	DMA synchronization events: Updated the information regarding the document reference.
19	Table 15, C6202/02B DSP Interrupts: Changed the document reference in the second footnote to: TMS320C6000 DSP Interrupt Selector Reference Guide (literature number SPRU646)
40	Figure 8, PWRD Field of the CSR Register: Updated the PD1, PD2, and PD3 information that follows Figure 8.
50	switching characteristics over recommended operating conditions for CLKOUT2 table: Removed NO. 1 (parameter $t_{\text{c(CKO2)}}$) from the table.





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PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TMS320C6202BGNY250	ACTIVE	FC/CSP	GNY	384	90	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202BGNY300	ACTIVE	FC/CSP	GNY	384	90	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202BGNZ250	ACTIVE	FCBGA	GNZ	352	40	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202BGNZ300	ACTIVE	FCBGA	GNZ	352	40	TBD	Call TI	Level-4-220C-72 HR
TMS320C6202BZNY250	ACTIVE	FC/CSP	ZNY	384	90	Pb-Free (RoHS Exempt)	SNAGCU	Level-4-260C-72HR
TMS320C6202BZNZ250	ACTIVE	FCBGA	ZNZ	352	40	Pb-Free (RoHS Exempt)	SNAGCU	Level-4-260C-72HR
TMS320C6202BZNZ300	ACTIVE	FCBGA	ZNZ	352	40	Pb-Free (RoHS Exempt)	SNAGCU	Level-4-260C-72HR
TMS320C6202GJL200	NRND	FCBGA	GJL	352	40	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GJL250	NRND	FCBGA	GJL	352	40	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GJL250X	NRND	FCBGA	GJL	352	40	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GJLA200	NRND	FCBGA	GJL	352	40	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GJLA233	NRND	FCBGA	GJL	352	40	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GLS200	NRND	FC/CSP	GLS	384	90	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GLS250	NRND	FC/CSP	GLS	384	90	TBD	SNPB	Level-4-220C-72 HR
TMS320C6202GLS250X	NRND	FC/CSP	GLS	384		TBD	Call TI	Call TI
TMS32C6202BGNZA250	ACTIVE	FCBGA	GNZ	352	40	TBD	SNPB	Level-4-220C-72 HR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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